

DDR4 SDRAM 8Gbit Datasheet

XCDJ512M16AP-QSNTX

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Using This Document

This document is intended for hardware and software engineer’s general information on the XCDJ512M16AP-QSNTX. Though every effort has been made to ensure that this document is current and accurate, more information may have become available subsequent to the production of this guide.

Revision History

Revision	Date	Description
Rev 1.0	2023/03/17	Create new document
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1. INTRODUCTION

XCDJ512M16AP-QSNTX is a 8G bits Double-Data-Rate-4 (DDR4) DRAMs by advanced Package technology.

XCDJ512M16AP-QSNTX offers space saving advantage that could miniaturize your portable device, and it is conformed with Green regulations.

1.1 Application

- Compact DSC / CAR Black Box / Action Cam / 360 Cam
- Drone
- Wearable

1.2 Features

- ◆ Power supply : VDD = VDDQ = 1.2V (1.14V to 1.26V); VPP = 2.5V (2.375V to 2.75V)
- ◆ JEDEC standard package: x16 96-ball FBGA
- ◆ Array Configuration : 8 banks (x16) 2 groups of 4 banks
- ◆ 8n-bit prefetch architecture
- ◆ Burst Length (BL): 8 and 4 with Burst Chop (BC)
- ◆ Programmable CAS Latency (CL)
- ◆ Programmable CAS Write Latency (CWL)
- ◆ Internal generated Vref for data inputs
- ◆ On-Die Termination (ODT) : Support Nominal, Park and Dynamic ODT
- ◆ Differential clock and data strobe inputs (CK_t, CK_c; DQS_t, DQS_c)
- ◆ Interface: 1.2V Pseudo Open Drain (POD) IO
- ◆ Per DRAM Addressability (PDA)
- ◆ Data Bus Inversion (DBI)
- ◆ Data Mask (DM) for write data
- ◆ Maximum Power Saving Mode (MPSM)
- ◆ LP ASR(Low Power Auto Self Refresh) mode is supported
- ◆ Asynchronous reset for power up
- ◆ Precharge: Auto precharge option for each burst access
- ◆ Operating case temperature : $0^{\circ}\text{C} \leq T_{\text{Case}} \leq 95^{\circ}\text{C}$
- ◆ Support auto-refresh and self-refresh mode
- ◆ Average Refresh Period:
 - $7.8\mu\text{s}$ at $0^{\circ}\text{C} \leq T_{\text{Case}} \leq 85^{\circ}\text{C}$
 - $3.9\mu\text{s}$ at $85^{\circ}\text{C} < T_{\text{Case}} \leq 95^{\circ}\text{C}$
- ◆ Fine granularity refresh 2x, 4x mode for smaller tRFC
- ◆ Programmable data strobe preambles

- ◆Command Address (CA) Parity is supported
- ◆Write Cyclic Redundancy Code (CRC) is supported
- ◆hPPR and sPPR are supported
- ◆Connectivity test mode (TEN) is supported
- ◆Gear Down Mode
- ◆Output driver calibration through ZQ pin (RZQ: 240ohm ± 1%)
- ◆JEDEC JESD-79-4 compliant
- ◆RoHS compliant

Note:

The functionality described and the timing specifications included in this datasheet are for the DLL Enabled mode of operation (normal operation), unless specifically stated otherwise.

1.3 Speed Bins

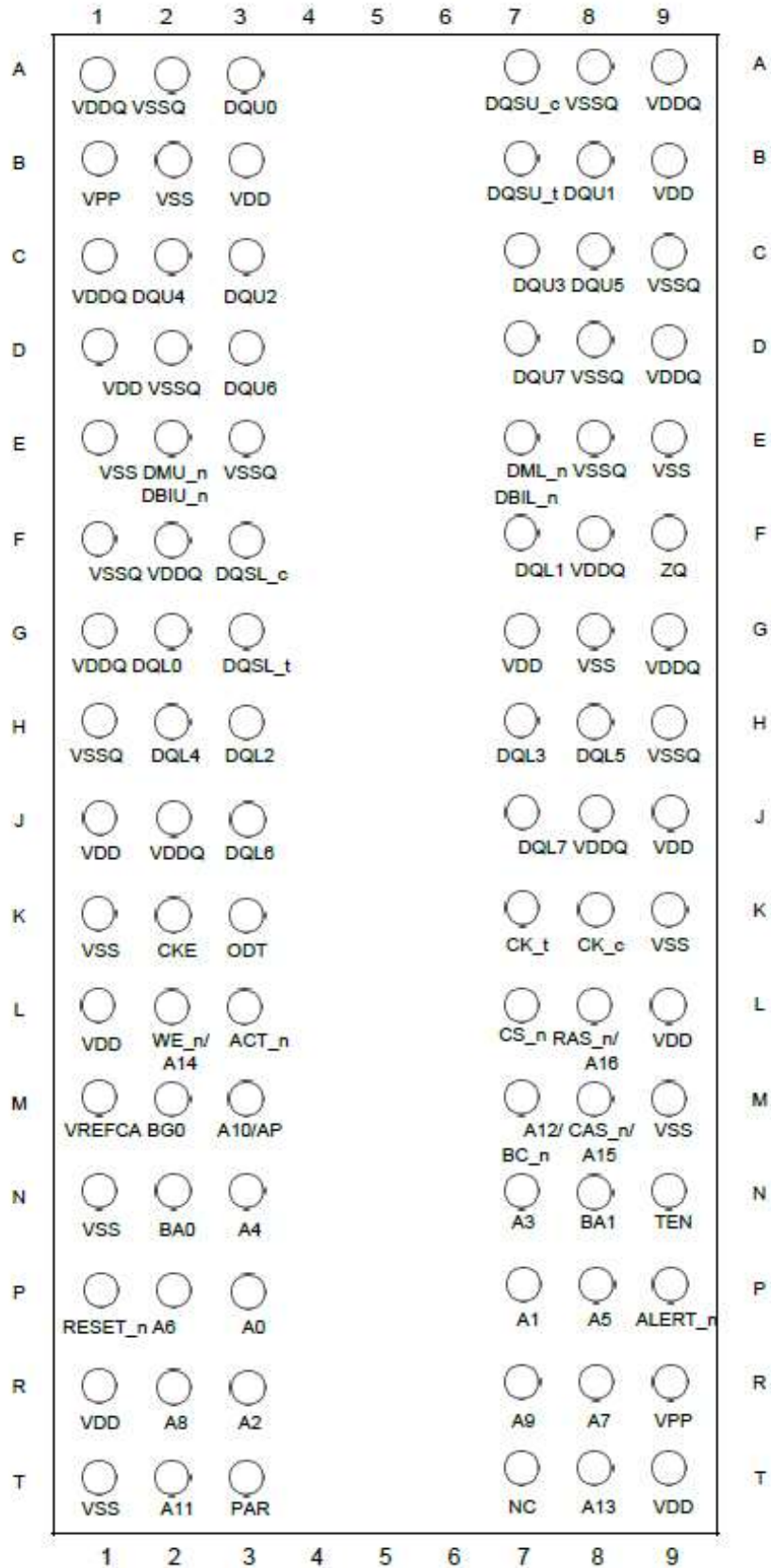
Speed	DDR4-3200	Unit
	22-22-22	
tCK (min)	0.625	ns
CAS Latency	22	nCK
tRCD (min)	13.75	ns
tRP (min)	13.75	ns
tRAS (min)	32	ns
tRC (min)	45.75	ns

1.4 Address Table

Param	512 Mb x16
Number of Bank Groups	2
Number of Banks per Bank Group	4
Bank Group Address	BG0
Bank Address per Bank Group	BA0~BA1
Row Address	A0~A15
Column Address	A0~A9
Page Size	2KB

2. Pin Configuration

2.1 Pin Assignment



TOP VIEW

2.2 Ball Description

Symbol	Type	Function
CK_t, CK_c	Input	Clock: CK_t and CK_c are differential clock inputs. All address and control input signals are sampled on the crossing of the positive edge of CK_t and negative edge of CK_c.
CKE	Input	Clock Enable: CKE High activates, and CKE Low deactivates, internal clock signals and device input buffers and output drivers. Taking CKE Low provides Precharge Power-Down and Self-Refresh operation (all banks idle), or Active Power-Down (row Active in any bank). CKE is synchronous for Self-Refresh exit. After VREFCA and Internal DQ Vref have become stable during the power on and initialization sequence, they must be maintained during all operations (including Self-Refresh). CKE must be maintained high throughout Read and Write accesses. Input buffers, excluding CK_t, CK_c, ODT and CKE are disabled during power-down. Input buffers, excluding CKE, are disabled during Self-Refresh.
CS_n	Input	Chip Select: All commands are masked when CS_n is registered HIGH. CS_n provides for external Rank selection on systems with multiple Ranks. CS_n is considered part of the command code.
ODT	Input	On Die Termination: ODT (registered High) enables RTT_NOM termination resistance internal to the DDR4 SDRAM. When enabled, ODT is only applied to each DQ, DQS_t, DQS_c and DM_n/DBI_n/TDQS_t, NU/TDQS_c (When TDQS is enabled via Mode Register A11=1 in MR1) signal for x8 configurations. For x16 configuration ODT is applied to each DQ, DQSU_t, DQSU_c, DQSL_t, DQSL_c, DMU_n, and DML_n signal. The ODT pin will be ignored if MR1
ACT_n	Input	Activation Command Input: ACT_n defines the Activation command being entered along with CS_n. The input into RAS_n/A16, CAS_n/A15 and WE_n/A14 will be considered as Row Address A16, A15 and A14.
RAS_n/A16, CAS_n/A15, WE_n/A14	Input	Command Inputs: RAS_n/A16, CAS_n/A15 and WE_n/A14 (along with CS_n) define the command being entered. These balls have multi function. For example, for activation with ACT_n Low, those are Addressing like A16,A15 and A14 but for non-activation command with ACT_n High, those are Command pins for Read, Write and other command defined in command truth table.
DM_n, DBI_n (DMU_n, DBIU_n DML_n, DBIL_n)	I/O	Input Data Mask and Data Bus Inversion: DM_n is an input mask signal for write data. Input data is masked when DM_n is sampled LOW coincident with that input data during a Write access. DM_n is sampled on both edges of DQS. DM is muxed with DBI function by Mode Register A10, A11, A12 setting in MR5. For x8 device, the function of DM or TDQS is enabled by Mode Register A11 setting in MR1. DBI_n is an input/output identifying whether to store/output the true or inverted data. If DBI_n is LOW, the data will be stored/output after inversion inside the DDR4 SDRAM and not inverted if DBI_n is HIGH. TDQS is
BG0-BG1	Input	Bank Group Inputs: BG0-BG1 define to which bank group an Active, Read, Write or Precharge command is being applied. BG0 also determines which mode register is to be accessed during a MRS cycle. x4/x8 have BG0 and BG1, but x16 has only BG0.

Symbol	Type	Function
BA0-BA1	Input	Bank Address Inputs: BA0-BA1 define to which bank an Active, Read, Write or Precharge command is being applied. Bank address also determines which mode register is to be accessed during a MRS cycle.
A0-A16	Input	Address Inputs: Provide the row address for ACTIVATE commands and the column address for Read/Write commands to select one location out of the memory array in the respective bank. (A10/AP, A12/BC_n, RAS_n/A16, CAS_n/A15 and WE_n/A14 have additional functions, see other rows. The address inputs also provide the op-code during Mode Register Set commands. A17 is only defined for the x4 configuration.
A10/AP	Input	Auto-precharge: A10 is sampled during Read/Write commands to determine whether Auto-precharge should be performed to the accessed bank after the Read/Write operation. (HIGH: Auto-precharge; LOW: no Auto-precharge). A10 is sampled during a Precharge command to determine whether the Precharge applies to one bank (A10 LOW) or all banks
A12/BC_n	Input	Burst Chop: A12/BC_n is sampled during Read and Write commands to determine if burst chop (on-the-fly) will be performed. (HIGH, no burst chop; LOW: burst chopped). See “Command Truth Table” of Operation Guide for details.
RESET_n	Input	Active Low Asynchronous Reset: Reset is active when RESET_n is LOW, and inactive when RESET_n is HIGH. RESET_n must be HIGH during normal operation. RESET_n is a CMOS rail to rail signal with DC high and low at 80% and 20% of VDD.
DQ	I/O	Data Input/Output: Bi-directional data bus. If CRC is enabled via mode register, then CRC code is added at the end of Data Burst. Any DQ from DQ3~DQ0 may indicate the internal Vref level during test via Mode Register Setting MR4 A4=High. During this mode, RTT value should be set to Hi-Z. Refer to vendor specific data sheets to determine which DQ is
DQS_t, DQS_c, DQSU_t, DQSU_c, DQSL_t, DQSL_c	I/O	Data Strobe: Output with Read data, input with Write data. Edge-aligned with Read data, centered-aligned with Write data. For x16, DQSL corresponds to the data on DQL0-DQL7; DQSU corresponds to the data on DQU0-DQU7. The data strobe DQS_t, DQSL_t and DQSU_t are paired with differential signals DQS_c, DQSL_c, and DQSU_c, respectively to provide differential pair signaling to the system during Reads and Writes. DDR4 SDRAM supports differential data strobe only and does not support single-ended.
TDQS_t, TDQS_c	Output	Termination Data Strobe: TDQS_t/TDQS_c is applicable for x8 DRAMs only. When enabled via Mode Register A11 = 1 in MR1, the DRAM will enable the same termination resistance function on TDQS_t/TDQS_c that is applied to DQS_t/DQS_c. When disabled via mode register A11 = 0 in MR1, DM/DBI/TDQS will provide the data mask function or Data Bus Inversion depending on MR5; A11, A12, A10 and TDQS_c is not used. x4/ x16 DRAMs must
PAR	Input	Command and Address Parity Input: DDR4 Supports Even Parity Check in DRAMs with MR setting. Once it is enabled via Register in MR5, then DRAM calculates Parity with ACT_n, RAS_n/A16, CAS_n/A15, WE_n/A14, BG0-BG1, BA0-BA1, A0-A16. Command and address inputs shall have parity check performed when commands are latched via the rising

Symbol	Type	Function
ALERT_n	I/O	ALERT: It has multi functions such as CRC error flag, Command and Address Parity error flag as output signal. If there is error in CRC, then Alert_n goes LOW for the period time interval and goes back HIGH. If there is error in Command Address Parity Check, then ALERT_n goes LOW for relatively long period until on going DRAM internal recovery
TEN	Input	Connectivity Test Mode Enable: Required on X16 devices and optional input on x4/x8 with densities equal to or greater than 8Gb. High in this pin will enable Connectivity Test Mode operation along with other pins. It is a CMOS rail to rail signal with AC high and low
NC		No Connect: No internal electrical connection is present.
VDDQ	Supply	DQ Power Supply: 1.2 V +/- 0.06 V
VSSQ	Supply	DQ Ground
VDD	Supply	Power Supply: 1.2 V +/- 0.06 V
VSS	Supply	Ground
VPP	Supply	DRAM Activating Power Supply: 2.5V (2.375V min, 2.75V max)
VREFCA	Supply	Reference voltage for CA
ZQ	Supply	Reference Pin for ZQ calibration

3 ABSOLUTE MAXIMUM RATINGS

3.1 Absolute Maximum DC Ratings

Table 3-1. Absolute Maximum DC Ratings

Symbol	Parameter	Min	Max	Unit	Note
VDD	Voltage on VDD pin relative to Vss	-0.3	1.5	V	1,3
VDDQ	Voltage on VDDQ pin relative to Vss	-0.3	1.5	V	1,3
VPP	Voltage on VPP pin relative to Vss	-0.3	3.0	V	4
V _{IN} , V _{OUT}	Voltage on any pin except VrefCA relative to Vss	-0.3	1.5	V	1,3,5
T _{STG}	Storage Temperature	-55	100	°C	1,2

Note:

- Stresses greater than those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.
- Storage Temperature is the case surface temperature on the center/top side of the DRAM. For the measurement conditions, please refer to JEDEC standard JESD51-2.
- VDD and VDDQ must be within 300 mV of each other at all times; and VREFCA must be not greater than 0.6 x VDDQ, When VDD and VDDQ are less than 500 mV; VREF may be equal to or less than 300 mV.
- VPP must be equal or greater than VDD/VDDQ at all times.
- Overshoot area above 1.5V is specified in [Section 6.3.5](#) and [Section 6.3.6](#).

3.2 Recommended DC Operating Conditions

Table 3-2. Recommended DC Operating Conditions

Symbol	Parameter	Ratings			Unit	Note
		Min	Typ.	Max		
VDD	Supply voltage	1.14	1.2	1.26	V	1,2,3
VDDQ	Supply voltage for output	1.14	1.2	1.26	V	1,2,3
VPP	Wordline supply voltage	2.375	2.5	2.75	V	3

Note:

- Under all conditions VDDQ must be less than or equal to VDD.
- VDDQ tracks with VDD. AC parameters are measured with VDD and VDDQ tied together.
- DC bandwidth is limited to 20MHz.

3.3 DRAM Component Operating Temperature Range

Table 3-3. Operating Temperature Range

Symbol	Parameter	Rating	Unit	Note
T _{OPER}	Normal Temperature Range	0~85	°C	1,2
	Extended Temperature Range	85~95	°C	1,3

Note:

- Operating Temperature T_{OPER} is the case surface temperature on the center/top side of the DRAM. For measurement conditions, please refer to the JEDEC document JESD51-2.

- 2 The Normal Temperature Range specifies the temperatures where all DRAM specifications will be supported. During operation, the DRAM case temperature must be maintained between 0 - 85°C under all operating conditions for the commercial offering; The industrial and automotive temperature offerings allow the case temperature to go below 0°C to -40°C.
- 3 Some applications require operation of the DRAM in the Extended Temperature Range between 85 °C and 95 °C case temperature. Full specifications are guaranteed in this range, but the following additional conditions apply:

- Refresh commands must be doubled in frequency, therefore reducing the Refresh interval tREFI to 3.9 μs.

It is also possible to specify a component with 1X refresh (tREFI to 7.8μs) in the Extended Temperature Range.

Please refer to the DIMM SPD for option availability.

- If Self-Refresh operation is required in the Extended Temperature Range, then it is mandatory to either use the Manual Self-Refresh mode with Extended Temperature Range capability (MR2 A6 = 0 and MR2 A7 = 1) or enable the optional Auto Self-Refresh mode (MR2 A6 = 1 and MR2 A7 = 1).

4 AC AND DC INPUT MEASUREMENT LEVELS

4.1 AC and DC Logic Input Levels for Single-ended Signals

Table 4-1. Single-ended AC and DC Input Levels for Command and Address

Symbol	Parameter	2666/3200		Unit	Note
		Min	Max		
VIH.CA(DC75)	DC input logic high	-	-	V	
VIL.CA(DC75)	DC input logic low	-	-	V	
VIH.CA(DC65)	DC input logic high	VREFCA +	VDD	V	
VIL.CA(DC65)	DC input logic low	VSS	VREFCA -0.065	V	
VIH.CA(AC100)	AC input logic high	-	-	V	1
VIL.CA(AC100)	AC input logic low	-	-	V	1
VIH.CA(AC90)	AC input logic high	VREF +0.09	Note 2	V	1
VIL.CA(AC90)	AC input logic low	Note 2	VREF - 0.09	V	1
VREFCA(DC)	Reference voltage for ADD, CMD	0.49*VDD	0.51*VDD	V	2,3

Note:

1. See "Overshoot and Undershoot Specifications"
2. The AC peak noise on VREFCA may not allow VREFCA to deviate from VREFCA(DC) by more than $\pm 1\%VDD$ (for reference: approx. $\pm 12mV$).
3. For reference: approx. $VDD/2 \pm 12 mV$.

4.2 AC and DC Logic Input Measurement Levels: Vref Tolerances

The DC-tolerance limits and AC-noise limits for the reference voltages VrefCA is illustrated in the Figure 4-1 below. It shows a valid reference voltage Vref(t) as a function of time. (Vref stands for VrefCA).

Vref(DC) is the linear average of Vref(t) over a very long period of time (for example, 1 second). This average has to meet the min/max requirement in Table 6-1. Furthermore Vref(t) may temporarily deviate from Vref(DC) by no more than $\pm 1\%$ VDD.

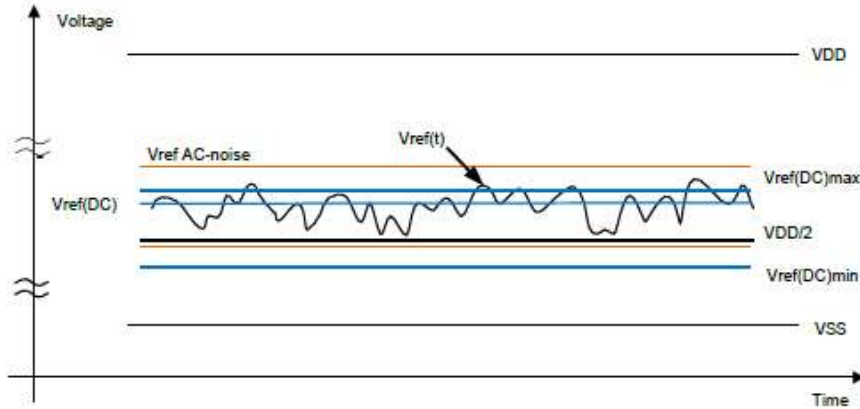


Figure 4-1. Illustration of Vref(DC) Tolerance and Vref AC-noise Limits

The voltage levels for setup and hold time measurements VIH(AC), VIH(DC), VIL(AC) and VIL(DC) are dependent on Vref. “Vref” should be understood as Vref(DC)

This clarifies that DC-variations of Vref affect the absolute voltage a signal has to reach to achieve a valid high or low level, and therefore the time to which setup and hold is measured. System timing and voltage budgets need to account for Vref(DC) deviations from the optimum position within the data-eye of the input signals.

This also clarifies that the DRAM setup/hold specification and derating values need to include time and voltage associated with Vref AC-noise. Timing and voltage effects due to AC-noise on Vref up to the specified limit ($\pm 1\%$ of VDD) are included in DRAM timings and their associated deratings.

4.3 AC and DC Logic Input Levels for Differential Signals

4.3.1 AC and DC Logic Input Levels for Differential Signals

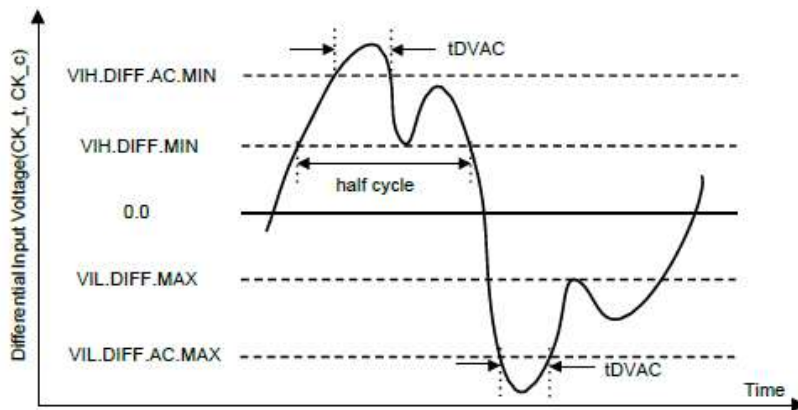


Figure 4-2. Definition of Differential AC-Swing and “Time above AC-Level” tDVAC

Note:

1. Differential signal rising edge from VIL.DIFF.MAX to VIH.DIFF.MIN must be monotonic slope.
2. Differential signal falling edge from VIH.DIFF.MIN to VIL.DIFF.MAX must be monotonic slope.

4.3.2 Differential Swing Requirements for Clock (CK_t - CK_c)

Table 4-2. Differential Input Levels Requirements for CK_t - CK_c

Symbol	Parameter	3200		Unit	Note
		Min	Max		
VIHdiff	differential input high	+ 0.110	Note 3	V	1
VILdiff	differential input low	Note 3	- 0.110	V	1
VIHdiff(AC)	differential input high ac	2 x(VIH(AC)-VREF)	Note 3	V	2
VILdiff(AC)	differential input low ac	Note 3	2x(VIL(AC) - VREF)	V	2

Note:

- Used to define a differential signal slew-rate.
- for CK_t - CK_c use VIH.CA/VIL.CA(AC) of ADD/CMD and VREFCA;
- These values are not defined; however, the differential signals CK_t - CK_c, need to be within the respective limits (VIH.CA(DC) max, VIL.CA(DC)min) for single-ended signals as well as the limitations for overshoot and undershoot.

Table4-3. Allowed Time before Ringback (tDVAC) for CK_t - CK_c

Slew Rate [V/ns]	tDVAC [ps] @ VIH/Ldiff(AC) = 200mV		tDVAC [ps] @ VIH/Ldiff(AC) = TBDmV	
	Min	Max	Min	Max
> 4.0	120	-	TBD	-
4.0	115	-	TBD	-
3.0	110	-	TBD	-
2.0	105	-	TBD	-
1.8	100	-	TBD	-
1.6	95	-	TBD	-
1.4	90	-	TBD	-
1.2	85	-	TBD	-
1.0	80	-	TBD	-
<1.0	80	-	TBD	-

4.3.3 Differential Swing Requirements for Clock (CK_t - CK_c)

Each individual component of a differential signal (CK_t, CK_c) has also to comply with certain requirements for single-ended signals.

CK_t and CK_c have to approximately reach VSEHmin/VSELmax (approximately equal to the ac-levels (VIH.CA(ac)/VIL.CA(ac)) for ADD/CMD signals) in every half-cycle.

Note that the applicable ac-levels for ADD/CMD might be different per speed-bin etc. E.g., if different value than VIH.CA(AC100)/VIL.CA(AC100) is used for ADD/CMD signals, then these ac-levels apply also for the single-ended signals CK_t and CK_c.

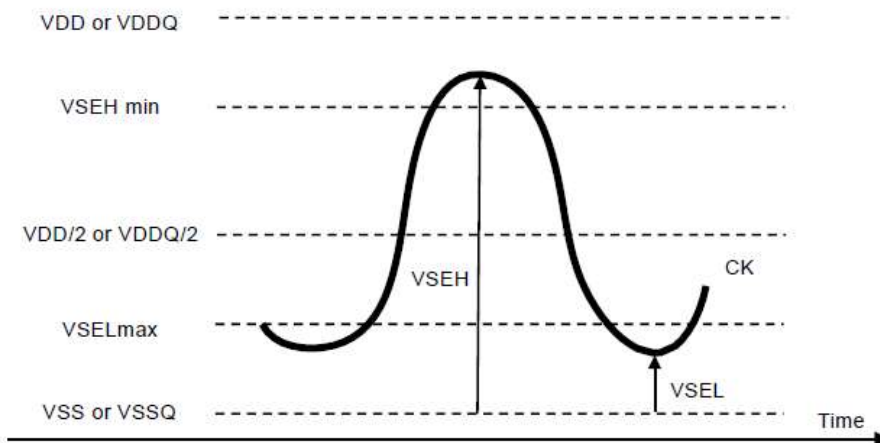


Figure 4-3. Single-ended Requirement for CK

Note that, while ADD/CMD signal requirements are with respect to VrefCA, the single-ended components of differential signals have a requirement with respect to VDD/2; this is nominally the same. The transition of single-ended signals through the ac-levels is used to measure setup time. For single-ended components of differential signals the requirement to reach VSELmax, VSEHmin has no bearing on timing, but adds a restriction on the common mode characteristics of these signals.

Table 4-4. Single-ended Levels Requirements for CK_t, CK_c

Symbol	Parameter	DDR4		Unit	Note
		Min	Max		
VSEH	Single-ended high-level for CK _t / CK _c	VDD/2 + 0.085	Note 3	V	1,2
VSEL	Single-ended low-level for CK _t / CK _c	Note 3	VDD/2 - 0.085	V	1,2

Note

1. For CK_t-CK_c use VIH.CA/VIL.CA(AC) of ADD/CMD.
2. VIH(AC)/VIL(AC) for ADD/CMD is based on VREFCA.
3. These values are not defined, however the single-ended signals CK_t, CK_c need to be within the respective limits (VIH.CA(DC) max, VIL.CA(DC)min) for single-ended signals as well as the limitations for overshoot and undershoot.

4.3.4 Address, Command, and Control Overshoot/Undershoot Specifications

Table 4-5. AC Overshoot/Undershoot Specification for Address, Command, and Control Pins

Parameter	Symbol	DDR4-3200	Unit	Note
Maximum peak amplitude above V_{AOS}	V_{AOSP}	0.06	V	
Upper boundary of overshoot area A_{AOS1}	V_{AOS}	$VDD + 0.24$	V	1
Maximum peak amplitude allowed for undershoot	V_{AUS}	0.3	V	
Maximum overshoot area per 1tCK Above V_{AOS}	A_{AOS2}	0.0055	V-ns	
Maximum overshoot area per 1tCK Between VDD and V_{AOS}	A_{AOS1}	0.1699	V-ns	
Maximum undershoot area per 1tCK Below VSS	A_{AUS}	0.1762	V-ns	

(A0-A13,A17,BG0-BG1,BA0-BA1,ACT_n,RAS_n/A16,CAS_n/A15,WE_n/A14,CS_n,CKE,ODT,C2-C0)

Note:

- The value of V_{AOS} matches VDD absolute max as defined in Table 3-1 if VDD equals VDD max as defined in Table 3-2. If VDD is above the recommended operating conditions, V_{AOS} remains at VDD absolute max as defined in Table 3-1.

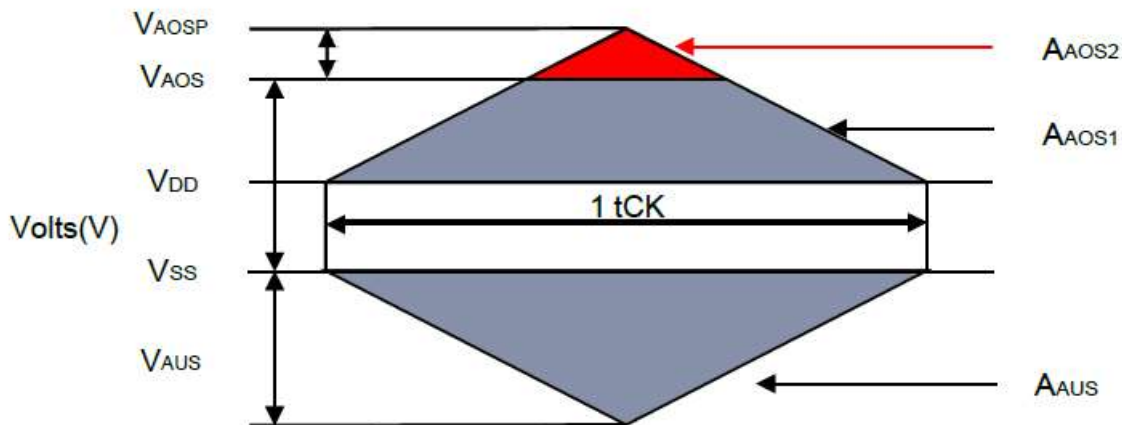


Figure 4-4. Address, Command, and Control Overshoot and Undershoot Definition

4.3.5 Address, Command, and Control Overshoot/Undershoot Specifications

Table 4-6. AC Overshoot/Undershoot Specification for Clock

Parameter	Symbol	2666/3200		Unit	Note
Maximum peak amplitude above	V_{COSP}	0.06		V	
Upper boundary of overshoot area	V_{COS}	$VDD + 0.24$		V	1
Maximum peak amplitude allowed for	V_{CUS}	0.3		V	
Maximum overshoot area per UI Above V_{COS}	A_{COS2}	0.0025	0.0025	V-ns	
Maximum overshoot area per 1tCK Between VDD and V_{COS}	A_{COS1}	0.075	0.075	V-ns	
Maximum undershoot area per UI Below VSS	A_{CUS}	0.0762	0.0762	V-ns	
(CK_t, CK_c)					

Note:

The value of VAOS matches VDD absolute max as defined in Table 3-1 if VDD equals VDD max as defined in Table 3-2. If VDD is above the recommended operating conditions, VAOS remains at VDD absolute max as defined in Table 3-1.

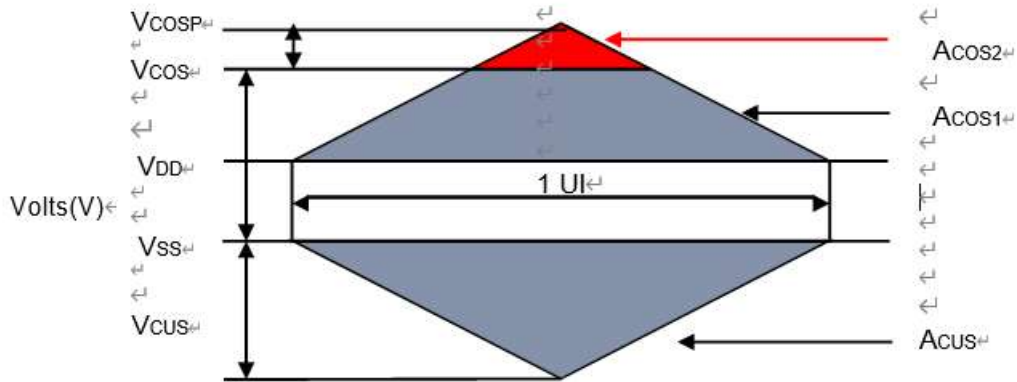


Figure 4-5. Clock Overshoot and Undershoot Definition

4.3.6 Data, Strobe and Mask Overshoot/Undershoot Specifications

Table 4-7. AC Overshoot/Undershoot Specification for Clock

Parameter	Symbol	2666/3200		Unit	Note
Maximum peak amplitude above V_{DOS}	V_{DOSP}	0.16		V	
Upper boundary of overshoot area A_{DOS1}	V_{DOS}	$VDD + 0.24$		V	1
Lower boundary of undershoot area A_{DUS1}	V_{DUS}	0.3		V	2
Maximum peak amplitude below V_{DUS}	A_{DUSP}	0.1		V-ns	
Maximum overshoot area per UI Above V_{DOS}	A_{DOS2}	0.0100	0.0100	V-ns	
Maximum overshoot area per 1UI Between V_{DDQ} and V_{DOS}	A_{DOS1}	0.0700	0.0700	V-ns	
Maximum undershoot area per UI Between V_{SSQ} and A_{DUS1}	A_{DUS1}	0.0700	0.0700	V-ns	
Maximum undershoot area per 1 UI below V_{DUS}	A_{DUS2}	0.100	0.100	V-ns	
(DQ, DQS_t, DQS_c, DM_n, DBI_n, TDQS_t, TDQS_c)					

Note:

1. The value of V_{DOS} matches (VIN, VOUT) max as defined in Table 3-1 if V_{DDQ} equals V_{DDQ} max as defined in Table 3-2. If V_{DDQ} is above the recommended operating conditions, V_{DOS} remains at (VIN, VOUT) max as defined in Table 3-1.

2. The value of V_{DUS} matches (VIN, VOUT) min as defined in Table 3-1.

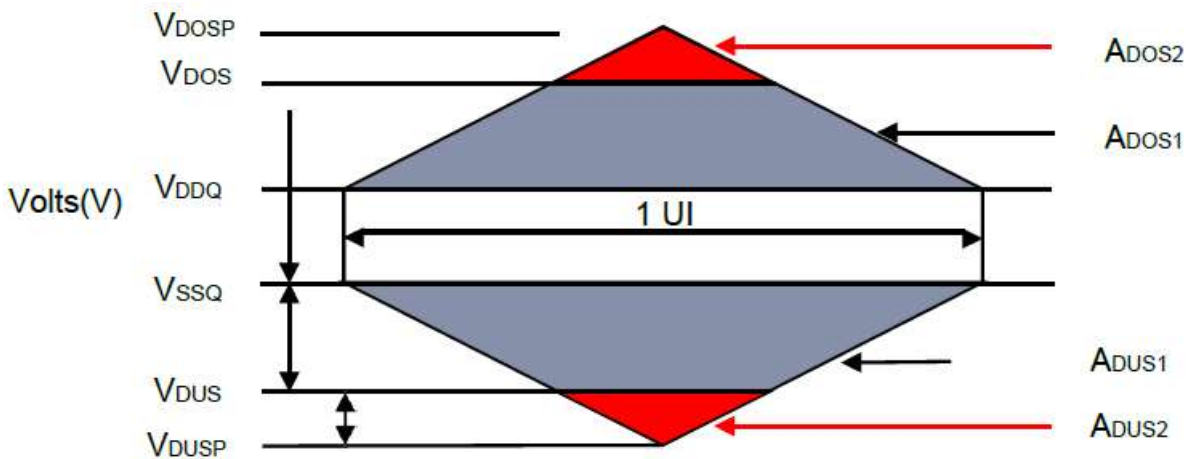


Figure 4-6. Data, Strobe and Mask Overshoot and Undershoot Definition

4.4 Slew Rate Definitions for Differential Input Signals

4.4.1 Slew Rate Definitions for Differential Input Signals

Input slew rate for differential signals (CK_t, CK_c) are defined and measured as shown in Table 4-8 and Figure 4-7.

Table 4-8. CK Differential Input Slew Rate Definition

Description	From	To	Defined by
Differential input slew rate for rising edge (CK_t-CK_c)	VILdiffmax	VIHdiffmin	$[VIHdiffmin - VILdiffmax] / \Delta TRdiff$
Differential input slew rate for falling edge (CK_t-CK_c)	VIHdiffmin	VILdiffmax	$[VIHdiffmin - VILdiffmax] / \Delta TFdiff$

Note:

- The differential signal (i. e., CK_t-CK_c) must be linear between these thresholds.

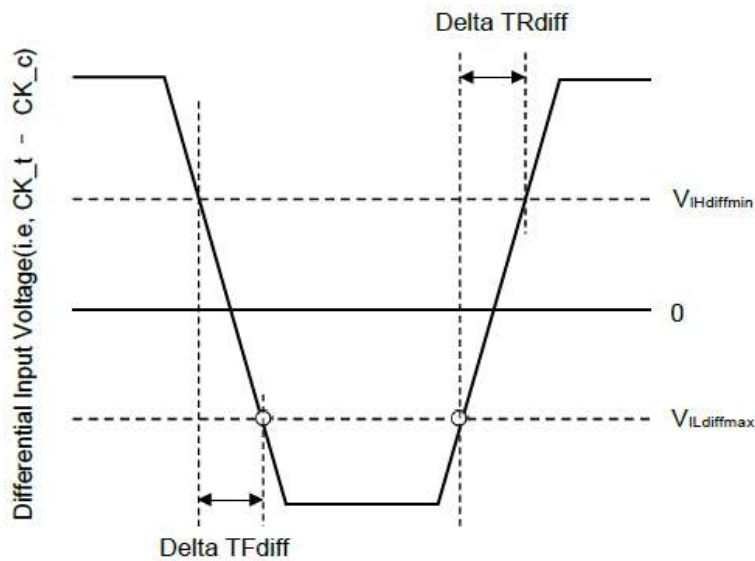


Figure 4-7. Differential Input Slew Rate Definition for CK_t, CK_c

4.4.2 Slew Rate Definitions for Differential Input Signals (CMD/ADD)

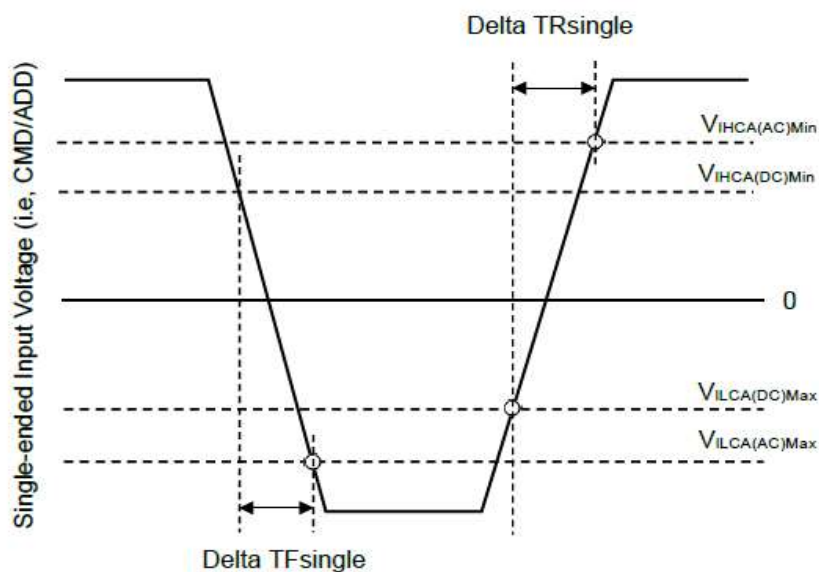


Figure 4-8. Single-ended Input Slew Rate Definition for CMD and ADD

Note:

1. Single-ended input slew rate for rising edge = $\{VIHCA(AC)Min - VILCA(DC)Max\} / \Delta TR$ single.
2. Single-ended input slew rate for falling edge = $\{VIHCA(DC)Min - VILCA(AC)Max\} / \Delta TF$ single.
3. Single-ended signal rising edge from $VILCA(DC)Max$ to $VIHCA(DC)Min$ must be monotonic slope.
4. Single-ended signal falling edge from $VIHCA(DC)Min$ to $VILCA(DC)Max$ must be monotonic slope.

4.5 CK Differential Input Point Voltage

To guarantee tight setup and hold times as well as output skew parameters with respect to clock, each cross point voltage of differential input signals (CK_t, CK_c) must meet the requirements in. The differential input cross point voltage V_{ix} is measured from the actual cross point of true and complement signals to the midlevel between of VDD and VSS.

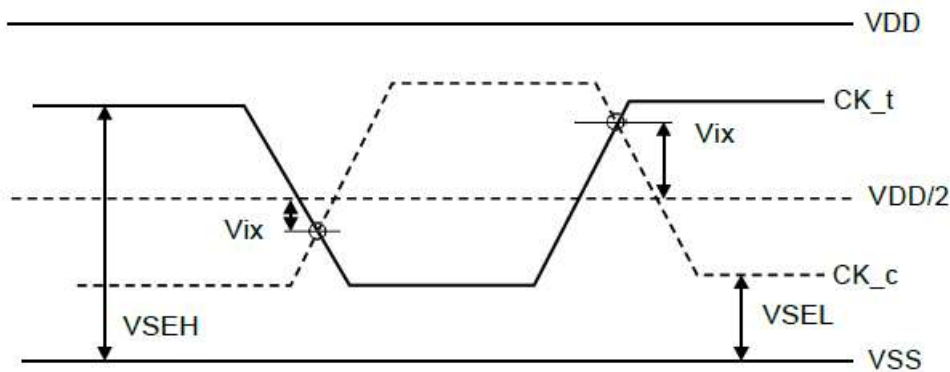


Figure 4-9. V_{ix} Definition (CK)

Table 4-10. Cross Point Voltage for CK Differential Input Signals at DDR4-3200

Symbol	Parameter	DDR4 – 3200			
		Min		Max	
		VSEL	VSEH	VSEL	VSEH
-	Area of VSEH,VSEL	$VSEL < VDD/2 - 145\text{mV}$	$VDD/2 - 145\text{mV} \leq VSEL \leq VDD/2 - 100\text{mV}$	$VDD/2 + 100\text{mV} \leq VSEL \leq VDD/2 + 145\text{mV}$	$VDD/2 + 145\text{mV} < VSEH$
VIX(CK)	Differential Input Cross Point Voltage relative to VDD/2 for CK_t, CK_c	-110mv	$-(VDD/2 - VSEL) + 30\text{mV}$	$(VSEH - VDD/2) - 30\text{mV}$	110mv

4.6 CMOS Rail to Rail Input Levels for RESET_n

Table 4-11. CMOS Rail to Rail Input Levels for RESET_n

Parameter	Symbol	Min	Max	Unit	Note
AC Input High Voltage	VIH(AC)_RESET	0.8*VDD	VDD	V	6
DC Input High Voltage	VIH(DC)_RESET	0.7*VDD	VDD	V	2
AC Input Low Voltage	VIL(AC)_RESET	VSS	0.2VCC	V	7
DC Input Low Voltage	VIL(DC)_RESET	VSS	0.3VCC	V	1
Rising Time	TR_RESET	-	1.0	μs	4
RESET Pulse Width	tPW_RESET	1.0	-	μs	3,5

Note:

1. After RESET_n is registered Low, RESET_n level shall be maintained below VIL(DC)_RESET during tPW_RESET, otherwise, SDRAM may not be reset.
2. Once RESET_n is registered High, RESET_n level must be maintained above VIH(DC)_RESET, otherwise, SDRAM operation will not be guaranteed until it is reset asserting RESET_n signal Low.
3. RESET is destructive to data contents.
4. No slope reversal (ringback) requirement during its level transition from Low to High.
5. This definition is applied only “Reset Procedure at Power Stable”.
6. Overshoot might occur. It should be limited by the Absolute Maximum DC Ratings.
7. Undershoot might occur. It should be limited by Absolute Maximum DC Ratings.

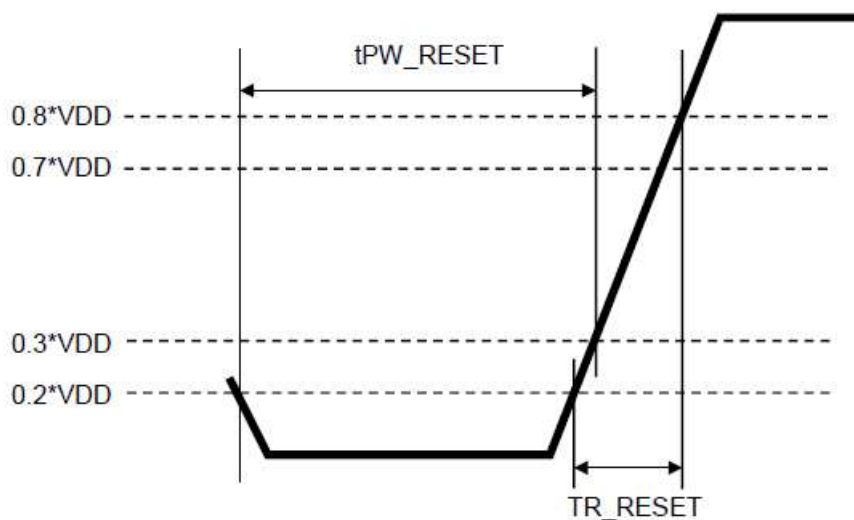


Figure 4-10. RESET_n Input Slew Rate Definition

4.7 AC&DC Logic Input Levels for DQS Signals

4.7.1 Differential Signal Definition

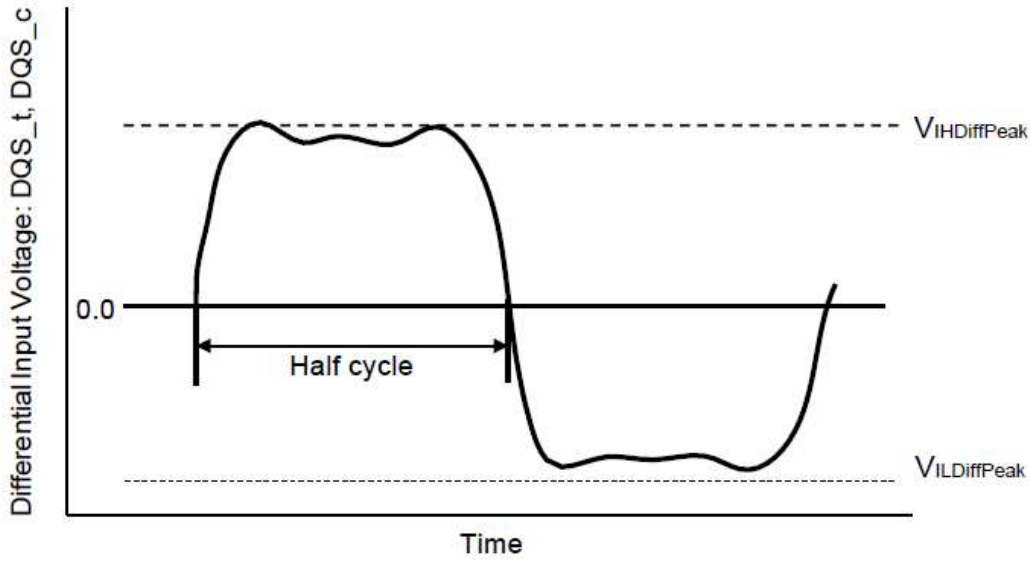


Figure4-11 DQS Differential Input Signal AC-swing Level

4.7.2 Differential Swing Requirements for DQS (DQS_t – DQS_c)

Table 4-12. Differential Input Swing Requirements for DQS

Symbol	Parameter	DDR4-3200		Unit	Note
		Min	Max		
V_IHDiffPeak	VIH.DIFF.Peak Voltage	140	Note2	mV	1
V_ILDiffPeak	VIL.DIFF.Peak Voltage	Note2	-140	mV	1

Note:

- Used to define a differential signal slew-rate.
- These values are not defined; however, the differential signals DQS_t - DQS_c, need to be within the respective limits of Overshoot, Undershoot Specification for single-ended signals.

4.7.3 Peak Voltage Calculation Method

The peak voltage of Differential DQS signals are calculated in a following equation.

$$VIH.DIFF.Peak\ Voltage = \text{Max}(f(t))$$

$$VIL.DIFF.Peak\ Voltage = \text{Min}(f(t))$$

$$f(t) = VDQS_t - VDQS_c$$

The $\text{Max}(f(t))$ or $\text{Min}(f(t))$ used to determine the midpoint which to reference the +/-35% window of the exempt non-monotonic signaling shall be the smallest peak voltage observed in all UI's.

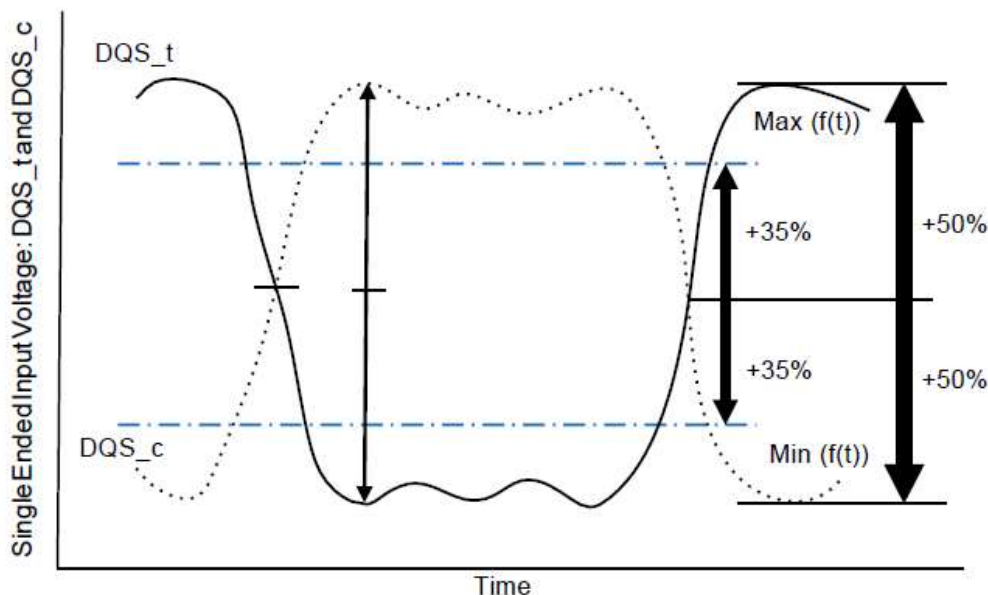


Figure 4-12. Definition of Differential DQS Peak Voltage and Range of Exempt Nonmonotonic Signaling

4.7.4 Differential Input Cross Point Voltage

To achieve tight RxMask input requirements as well as output skew parameters with respect to strobe, the cross point voltage of differential input signals (DQS_t, DQS_c) must meet the requirements in Table 4-13. The differential input cross point voltage VIX_DQS (VIX_DQS_FR and VIX_DQS_RF) is measured from the actual cross point of DQS_t, DQS_c relative to the VDQSmid of the DQS_t and DQS_c signals.

VDQSmid is the midpoint of the minimum levels achieved by the transitioning DQS_t and DQS_c signals, and noted by VDQS_trans. VDQS_trans is the difference between the lowest horizontal tangent above VDQSmid of the transitioning DQS signals and the highest horizontal tangent below VDQSmid of the transitioning DQS signals.

A nonmonotonic transitioning signal's ledge is exempt or not used in determination of a horizontal tangent provided the said ledge occurs within +/- 35% of the midpoint of either VIH.DIFF. Peak Voltage (DQS_t rising) or VIL.DIFF. Peak Voltage (DQS_c rising), refer to Figure 4-12. A secondary horizontal tangent resulting from a ring-back transition is also exempt in determination of a horizontal tangent. That is, a falling transition's horizontal tangent is derived from its negative slope to zero slope transition (point A in Figure 4-13) and a ring-back's horizontal tangent derived from its positive slope to zero slope transition (point B in Figure 4-13) is not a valid horizontal tangent; and a rising transition's horizontal tangent is derived from its positive slope to zero slope transition (point C in Figure 4-13) and

a ring-back's horizontal tangent derived from its negative slope to zero slope transition(point D in Figure 4-13) is not a valid horizontal tangent.

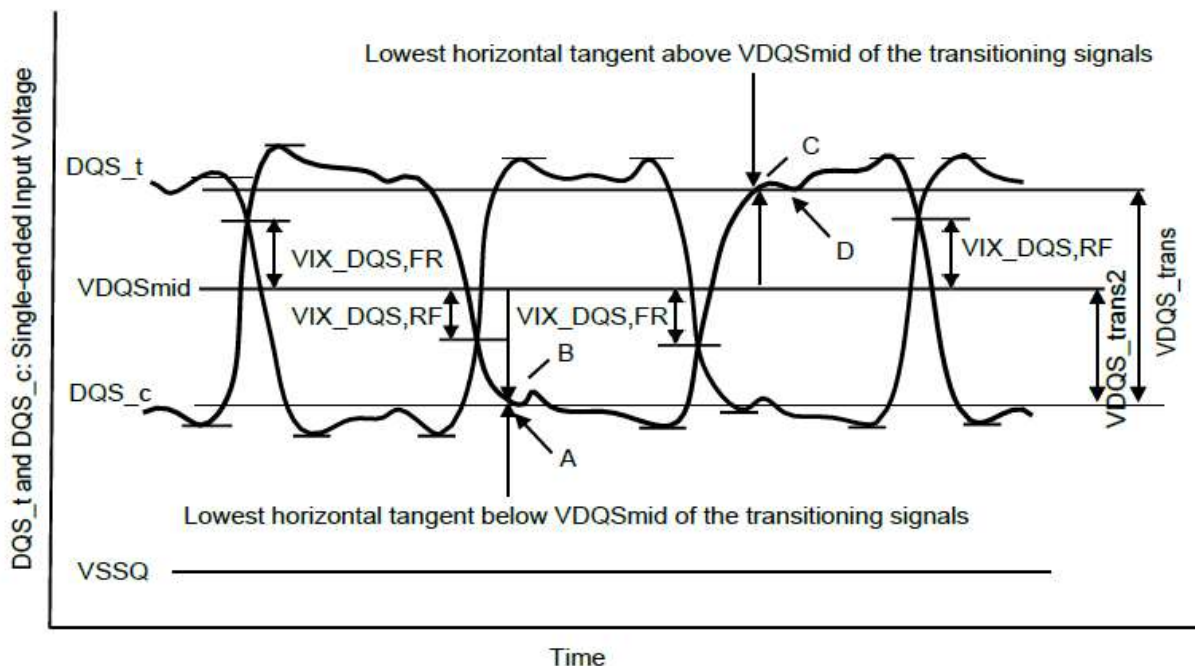


Figure 4-13. Vix Definition (DQS)

Table 4-13. Cross Point Voltage for DQS Differential Input Signals

Symbol	Parameter	DDR4-3200		Unit	Note
		Min	Max		
Vix_DQS_ratio	DQS Differential input crosspoint voltage ratio	-	25	%	1,2
VDQSmid_to_Vcent	VDQSmid offset relative to	-	min(VIHdiff,50)	mV	3,4,5

Note:

1. Vix_DQS_Ratio is DQS Vix crossing (Vix_DQS_FR or Vix_DQS_RF) divided by VDQS_trans. VDQS_trans is the difference between the lowest horizontal tangent above VDQSmid of the transitioning DQS signals and the highest horizontal tangent below VDQSmid of the transitioning DQS signals.
2. VDQSmid will be similar to the VREFDQ internal setting value obtained during Vref Training if the DQS and DQs drivers and paths are matched.
3. The maximum limit shall not exceed the smaller of VIHdiff minimum limit or 50mV.
4. Vix measurements are only applicable for transitioning DQS_t and DQS_c signals when toggling data, preamble and high-z states are not applicable conditions.
5. The parameter VDQSmid is defined for simulation and ATE testing purposes, it is not expected to be tested in a system.

4.7.5 Differential Input Slew Rate Definition

Input slew rate for differential signals (DQS_t, DQS_c) are defined and measured as shown in Figure 4-14 and Table 4-14.

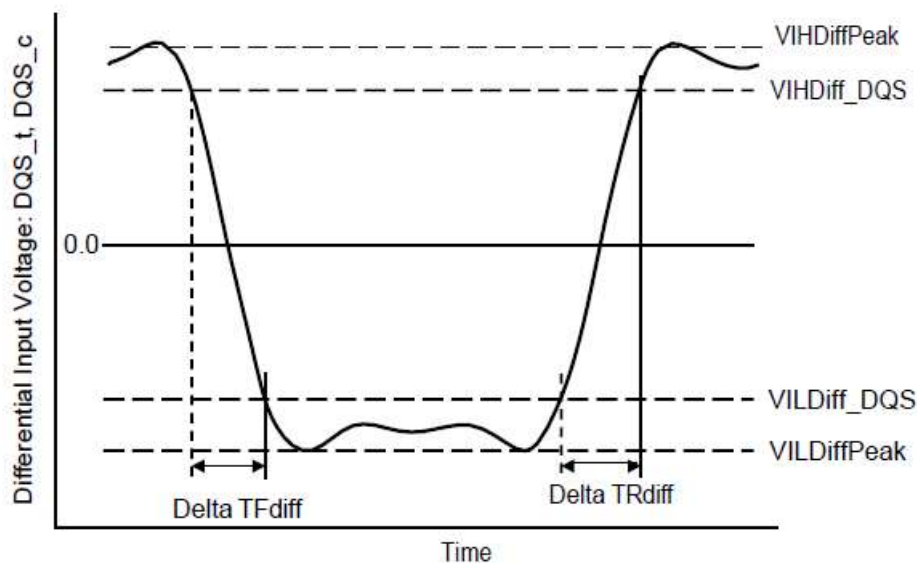


Figure 4-14. Differential Input Slew Rate Definition for DQS_t, DQS_c

Note:

1. Differential signal rising edge from VILDiff_DQS to VIHDiff_DQS must be monotonic slope.
2. Differential signal falling edge from VIHDiff_DQS to VILDiff_DQS must be monotonic slope.

Table 4-14. Differential Input Slew Rate Definition for DQS_t, DQS_c

Description	From	To	Defined by
Differential input slew rate for rising edge(DQS_t – DQS_c)	VILdiff_DQS	VIHdiff_DQS	$ VILDiff_DQS - VIHDiff_DQS /DeltaTRdiff$
Differential input slew rate for falling edge(DQS_t – DQS_c)	VIHdiff_DQS	VILdiff_DQS	$ VILDiff_DQS - VIHDiff_DQS /DeltaTRdiff$

Table 4-15. Differential Input Level for DQS_t, DQS_c

Symbol	Parameter	DDR4-3200		Unit
		Min	Max	
VIHdiff_DQS	DC input logic high	130	-	mV
VILdiff_DQS	DC input logic low	-	-130	mV

Table 4-16. Differential Input Slew Rate for DQS_t, DQS_c

Symbol	Parameter	DDR4-3200		Unit
		Min	Max	
SRldiff	Differential input slew rate	2.5	18	V/ns

5 AC&DC OUTPUT MEASUREMENT LEVELS

5.1 Output Driver DC Electronic Characteristics

The DDR4 driver supports two different RON values. These RON values are referred as strong (low RON) and weak mode (highRON). A functional representation of the output buffer is shown in Figure 5-1 below. Output driver impedance RON is defined as the individual pull-up and pull-down resistors (RON_{Pu} and RON_{Pd}).

$$RON_{Pu} = \frac{V_{DDQ} - V_{out}}{|I_{out}|} \text{ under the condition that } RON_{Pd} \text{ is off}$$

$$RON_{Pd} = \frac{V_{out}}{|I_{out}|} \text{ under the condition that } RON_{Pu} \text{ is off.}$$

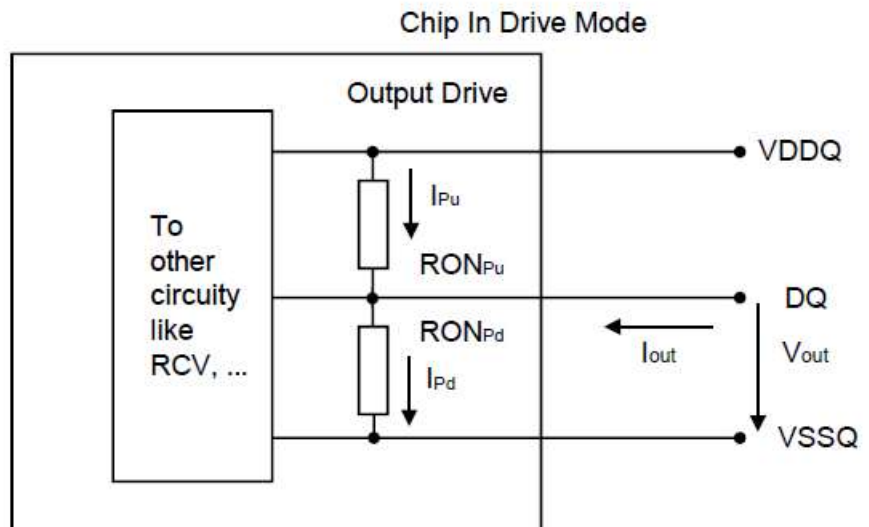


Figure 5-1. Output Driver

Table 5-1. Output Driver DC Electrical Characteristics, Assuming RZQ = 240ohm; Entire Operating Temperature Range; after Proper ZQ Calibration

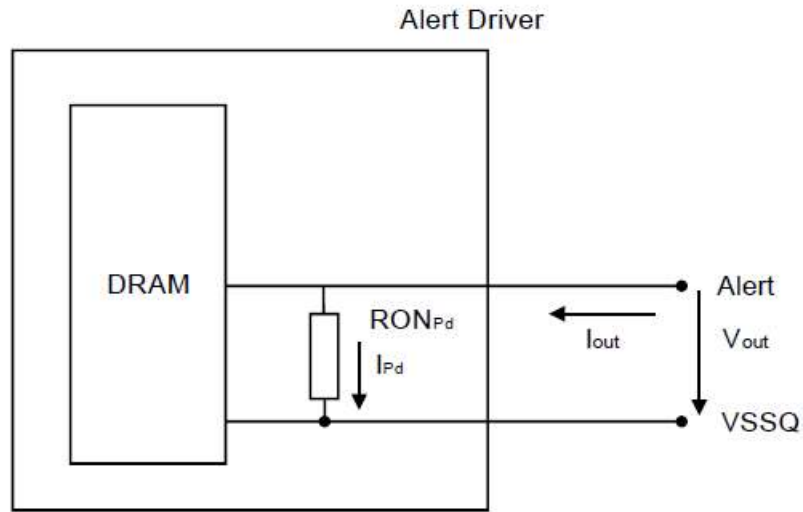
RONNOM	Resistor	Vout	Min	Nom	Max	Unit	Note
34Ω	RON34Pd	VOLdc=0.5*VDDQ	0.73	1.0	1.1	RZQ/7	1,2
		VOMdc=0.8*VDDQ	0.83	1.0	1.1	RZQ/7	1,2
		VOHdc=1.1*VDDQ	0.83	1.0	1.25	RZQ/7	1,2
	RON34Pu	VOLdc=0.5*VDDQ	0.9	1.0	1.1	RZQ/7	1,2
		VOMdc=0.8*VDDQ	0.9	1.0	1.1	RZQ/7	1,2
		VOHdc=1.1*VDDQ	0.8	1.0	1.25	RZQ/7	1,2
48Ω	RON48Pd	VOLdc=0.5*VDDQ	0.73	1.0	1.1	RZQ/5	1,2
		VOMdc=0.8*VDDQ	0.83	1.0	1.1	RZQ/5	1,2
		VOHdc=1.1*VDDQ	0.83	1.0	1.25	RZQ/5	1,2
	RON48Pu	VOLdc=0.5*VDDQ	0.9	1.0	1.25	RZQ/5	1,2
		VOMdc=0.8*VDDQ	0.9	1.0	1.1	RZQ/5	1,2
		VOHdc=1.1*VDDQ	0.8	1.0	1.1	RZQ/5	1,2
Mismatch between pull-up and pull-down, MMPuPd	VOMdc=0.8*VDDQ	-10		17	%	1,2,3,4	
Mismatch DQ-DQ within byte variation pull-up, MMPudd	VOMdc=0.8*VDDQ			10	%	1,2,4	
Mismatch DQ-DQ within byte variation pull-up, MMPddd	VOMdc=0.8*VDDQ			10	%	1,2,4	

Note:

- The tolerance limits are specified after calibration with stable voltage and temperature. For the behavior of the tolerance limits if temperature or voltage changes after calibration, see following section on voltage and temperature sensitivity(TBD).
- Pull-up and pull-dn output driver impedances are recommended to be calibrated at 0.8*VDDQ. Other calibration schemes may be used to achieve the linearity spec shown above, e.g. calibration at 0.5*VDDQ and 1.1*VDDQ.
- Measurement definition for mismatch between pull-up and pull-down, MMPuPd: Measure RONPu and RONPD both at 0.8*VDD separately; RONnom is the nominal RON value.
- $MMPuPdd = [(RONPu - RONPd / RONNOM)] * 100$
- RON variance range ratio to RON Nominal value in a given component, including DQS_t and DQS_c.
- $MMPuPdd = [(RONPuMax - RONPuMin / RONNOM)] * 100$
- $MMPdPdd = [(RONPdMax - RONPdMin / RONNOM)] * 100$
- This parameter of x16 device is specified for Upper byte and Lower byte.

5.1.1 Alert_n Output Driver Characteristic

A functional representation of the output buffer is shown in the figure below. Output driver impedance RON is defined as follows: $RON_{Pd} = \frac{V_{out}}{|I_{out}|}$ under the condition that RON_{Pu} is off.



Resister	Vout	Min	Max	Unit	Note
RON _{Pd}	$VOL_{dc}=0.1*VDDQ$	0.3	1.2	34Ω	1
	$VOM_{dc}=0.8*VDDQ$	0.3	1.2	34Ω	1
	$VOH_{dc}=1.1*VDDQ$	0.4	1.4	34Ω	1

Note:

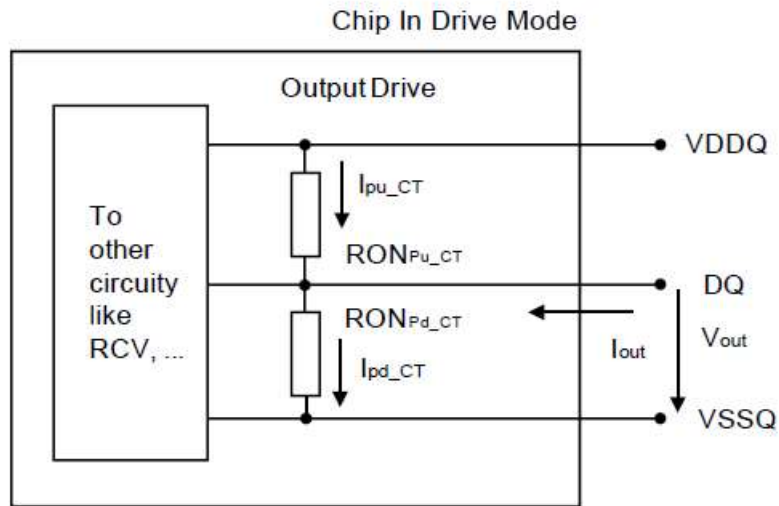
VDDQ voltage is at VDDQ DC. VDDQ DC definition is TBD.

5.1.2 Output Driver Characteristic of Connectivity Test (CT) Mode

Following output driver impedance RON will be applied Test Output Pin during Connectivity Test (CT) Mode. The individual pull-up and pull-down resistors (RONPu_CT and RONPd_CT) are defined as follows:

$$RON_{Pu_CT} = \frac{VDDQ - V_{out}}{|I_{out}|}$$

$$RON_{Pd_CT} = \frac{V_{out}}{|I_{out}|}$$



RON _{NOM_CT}	Resister	Vout	Max	Unit
34Ω	RON _{Pd_CT}	VOB _{dc} =0.2*V _{DDQ}	1.9	34Ω
		VOL _{dc} =0.5*V _{DDQ}	2.0	34Ω
		VOM _{dc} =0.8*V _{DDQ}	2.2	34Ω
		VOH _{dc} =1.1*V _{DDQ}	2.5	34Ω
	RON _{Pu_CT}	VOB _{dc} =0.2*V _{DDQ}	2.5	34Ω
		VOL _{dc} =0.5*V _{DDQ}	2.2	34Ω
		VOM _{dc} =0.8*V _{DDQ}	2.0	34Ω
		VOH _{dc} =1.1*V _{DDQ}	1.9	34Ω

Note:

Connectivity test mode uses un-calibrated drivers, showing the full range over PVT. No mismatch between pull up and pull down is defined.

5.2 Single-ended AC& DC Output Levels

Table 5-2. Single-ended AC&DC Output Levels

Symbol	Parameter	DDR4-3200	Unit
$V_{OH(DC)}$	DC output high measurement level (for IV curve linearity)	$1.1 \times V_{DDQ}$	V
$V_{OM(DC)}$	DC output mid measurement level (for IV curve linearity)	$0.8 \times V_{DDQ}$	V
$V_{OL(DC)}$	DC output low measurement level (for IV curve linearity)	$0.5 \times V_{DDQ}$	V
$V_{OH(AC)}$	AC output high measurement level (for output SR)	$(0.7+0.15) \times V_{DDQ}$	V
$V_{OL(AC)}$	AC output low measurement level (for output SR)	$(0.7-0.15) \times V_{DDQ}$	V

Note:

The swing of $\pm 0.15 \times V_{DDQ}$ is based on approximately 50% of the static single-ended output peak-to-peak swing with a driver impedance of $RZQ/7\Omega$ and an effective test load of 50Ω to $V_{TT} = V_{DDQ}$.

5.3 Differential AC&DC Output Levels

Table 5-3. Differential AC&DC Output Levels

Symbol	Parameter	DDR4-2666/DDR4-3200	Unit
$V_{OHdiff(AC)}$	AC differential output high measurement level (for output SR)	$+ 0.3 \times V_{DDQ}$	V
$V_{OLdiff(AC)}$	AC differential output low measurement level (for output SR)	$-0.3 \times V_{DDQ}$	V

Note:

The swing of $\pm 0.3 \times V_{DDQ}$ is based on approximately 50% of the static differential output peak-to-peak swing with a driver impedance of $RZQ/7\Omega$ and an effective test load of 50Ω to $V_{TT} = V_{DDQ}$ at each of the differential outputs.

5.4 Single-ended Output Slew Rate

With the reference load for timing measurements, output slew rate for falling and rising edges is defined and measured between $V_{OL}(AC)$ and $V_{OH}(AC)$ for single ended signals as shown in Table 5-4 and Figure 5-2.

Table 5-4. Single-ended Output Slew Rate Definition

Description	Measured		Defined by
	From	To	
Single ended output slew rate for rising edge	$V_{OL}(AC)$	$V_{OH}(AC)$	$[V_{OH}(AC) - V_{OL}(AC)] / \Delta TR_{se}$
Single ended output slew rate for falling edge	$V_{OH}(AC)$	$V_{OL}(AC)$	$[V_{OH}(AC) - V_{OL}(AC)] / \Delta TF_{se}$

Note:

Output slew rate is verified by designed and characterization, and may not be subject to production test.

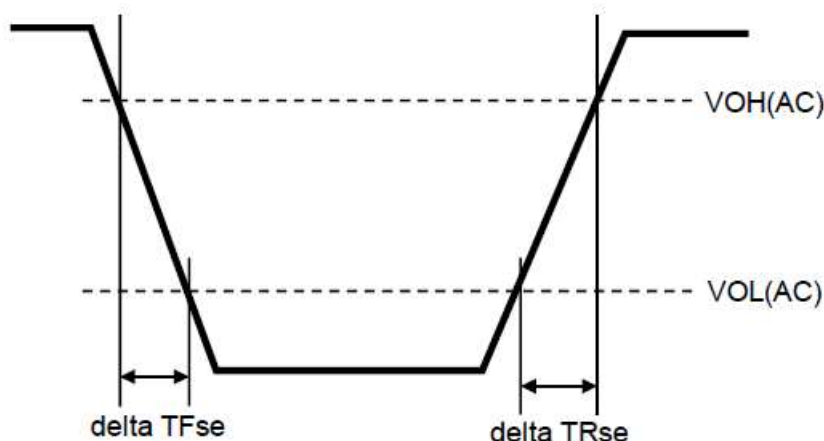


Figure 5-2. Single-ended Output Slew Rate Definition

Table 5-5. Single-ended Output Slew Rate

Parameter	Symbol	DDR4-3200		Unit
		Min	Max	
Single ended output slew rate	SRQ _{se}	4	9	V/ns

Description:

SR: Slew Rate; Q: Query Output (like in DQ, which stands for Data-in, Query-Output) se: Single-ended Signals;
For RON = RZQ/7 setting

Note:

In two cases, a maximum slew rate of 12 V/ns applies for a single DQ signal within a byte lane.

- Case 1 is defined for a single DQ signal within a byte lane which is switching into a certain direction (either from high to low or low to high) while all remaining DQ signals in the same byte lane are static (i.e. they stay at either high or low).
- Case 2 is defined for a single DQ signal within a byte lane which is switching into a certain direction (either from high to low or low to high) while all remaining DQ signals in the same byte lane are switching into the opposite direction (i.e. from low to high or high to low respectively). For the remaining DQ signal switching into the opposite direction, the regular maximum limit of 9 V/ns applies).

5.5 Differential Output Slew Rate

With the reference load for timing measurements, output slew rate for falling and rising edges is defined and measured between VOLdiff(AC) and VOHdiff(AC) for differential signals as shown in Table5-6 and Figure 5-3.

Table 5-6. Differential Output Slew Rate Definition

Description	Measured		Defined by
	From	To	
Differential output slew rate for rising	VOLdiff(AC)	VOHdiff(AC)	$[V_{OHdiff(AC)} - V_{OLdiff(AC)}] / \Delta TR_{diff}$
Differential output slew rate for falling	VOHdiff(AC)	VOLdiff(AC)	$[V_{OHdiff(AC)} - V_{OLdiff(AC)}] / \Delta TF_{diff}$

Note:

Output slew rate is verified by design and characterization, and may not be subject to production test.

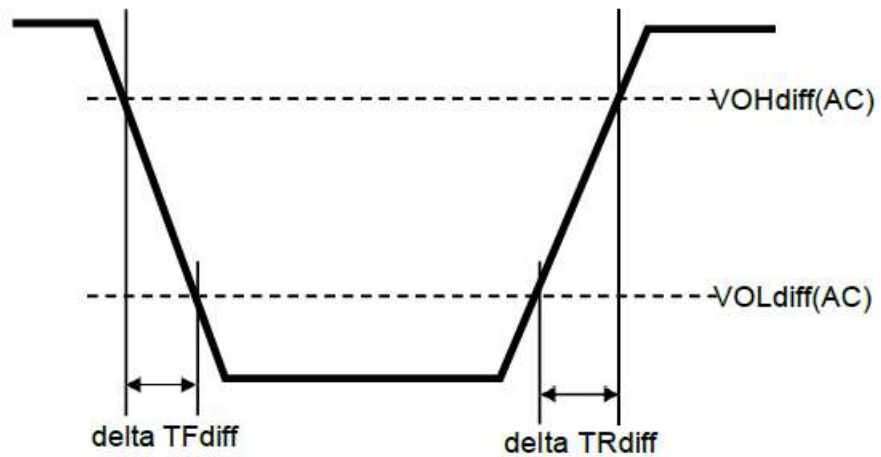


Figure 5-3. Differential Output Slew Rate Definition

Table 5-7. Differential Output Slew Rate

Parameter	Symbol	DDR4-3200		Unit
		Min	Max	
Differential output slew rate	SRQdiff	8	18	V/ns

Description:

SR: Slew Rate; Q: Query Output (like in DQ, which stands for Data-in, Query-Output); Diff: Differential Signals; For RON = RZQ/7 setting.

5.6 Single-ended AC& DC Output Levels of Connectivity Test Mode

Following output parameters will be applied for DDR4 SDRAM Output Signal during Connectivity Test Mode.

Table 5-8. Single-ended AC&DC Output Level of Connectivity Test Mode

Symbol	Parameter	DDR4-3200	Unit	Note
$V_{OH(DC)}$	DC output high measurement level (for IV curve linearity)	$1.1 \times V_{DDQ}$	V	
$V_{OM(DC)}$	DC output mid measurement level (for IV curve linearity)	$0.8 \times V_{DDQ}$	V	
$V_{OL(DC)}$	DC output low measurement level (for IV curve linearity)	$0.5 \times V_{DDQ}$	V	
$V_{OB(DC)}$	DC output below measurement level (for IV curve linearity)	$0.2 \times V_{DDQ}$	V	
$V_{OH(AC)}$	AC output high measurement level (for output SR)	$V_{TT} + (0.1 \times V_{DDQ})$	V	1
$V_{OL(AC)}$	AC output below measurement level (for output SR)	$V_{TT} - (0.1 \times V_{DDQ})$	V	1

Note:

The effective test load is 50Ω terminated by $V_{TT} = 0.5 \times V_{DDQ}$.

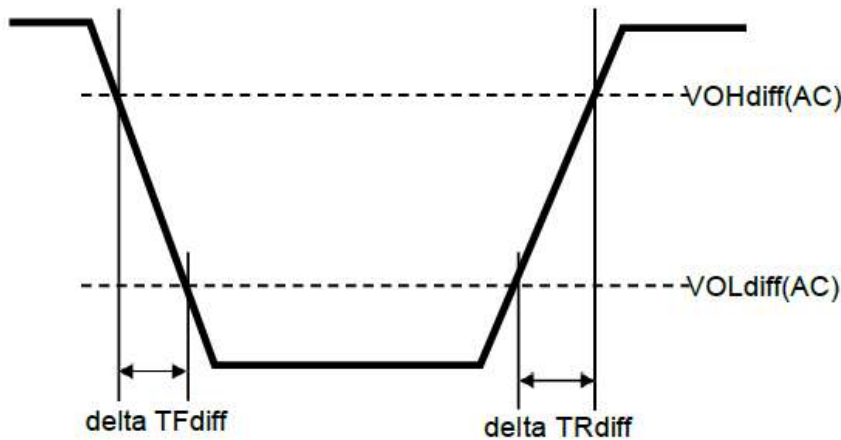


Figure 5-4. Output Slew Rate Definition of Connectivity Test Mode

Table 5-9. Single-ended Output Slew Rate of Connectivity Test Mode

Parameter	Symbol	DDR4-3200		Unit
		Min	Max	
Output signal Falling time	TF_output_CT	-	10	ns/V
Output signal Rising time	TR_output_CT	-	10	ns/V

5.7 Reference Load for Connectivity Test Mode Timing

The reference load for ODT timings is defined in Figure 5-5

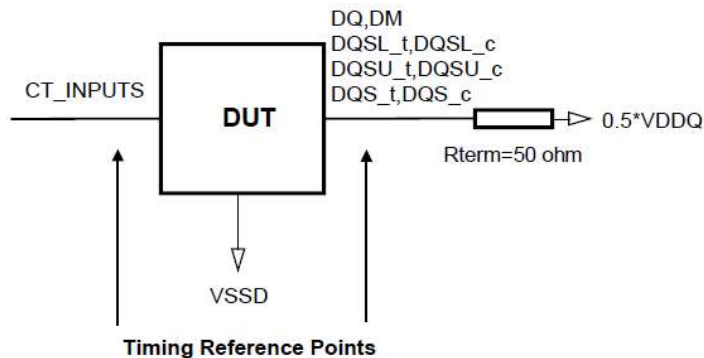


Figure 5-5. Connectivity Test Mode Timing Reference Load

6 SPEED BIN

6.1 DDR4-3200 Speed Bins and Operations

Table 6-1. DDR4-3200 Speed Bins and Operations

Speed Bin			DDR4-3200		Unit	Note
CL-nRCD-nRP			22-22-22			
Parameter	Symbol		Min	Max		
Internal Read command to first data	tAA		13.75	18.00		12
Internal Read command to first data with	tAA_DBI		tAA(min) + 4nCK	tAA(max) + 4nCK		12
ACT to internal Read or Write delay time	tRCD		13.75	-		12
PRE command period	tRP		13.75	-		12
ACT to PRE command period	tRAS		32	9 x tREFI		12
ACT to ACT or REF command period	tRC		45.75	-		12
	Normal	Read DBI				
CWL = 9	CL = 9	CL = 11	tCK(AVG)	Reserved		1,2,3,4,9,11
	CL = 10	CL = 12	tCK(AVG)	Reserved		1,2,3,9,11
CWL = 9,11	CL = 10	CL = 12	tCK(AVG)	Reserved		4
	CL = 11	CL = 13	tCK(AVG)	1.25	<1.5	1,2,3,4,9
	CL = 12	CL = 14	tCK(AVG)	1.25	<1.5	1,2,3,9
CWL = 10,12	CL = 12	CL = 14	tCK(AVG)	Reserved		4
	CL = 13	CL = 15	tCK(AVG)	1.071	<1.25	1,2,3,4,9
	CL = 14	CL = 16	tCK(AVG)	1.071	<1.25	1,2,3,9
CWL = 11,14	CL = 14	CL = 17	tCK(AVG)	Reserved		4
	CL = 15	CL = 18	tCK(AVG)	0.937	<1.071	1,2,3,4,9
	CL = 16	CL = 19	tCK(AVG)	0.937	<1.071	1,2,3,9
CWL = 12,16	CL = 15	CL = 18	tCK(AVG)	Reserved		4
	CL = 16	CL = 19	tCK(AVG)	Reserved		1,2,3,4,9
	CL = 17	CL = 20	tCK(AVG)	0.833	<0.937	1,2,3,4,9
	CL = 18	CL = 21	tCK(AVG)	0.833	<0.937	1,2,3,9
CWL = 14,18	CL = 17	CL = 20	tCK(AVG)	Reserved		1,2,3,4
	CL = 18	CL = 21	tCK(AVG)	Reserved		1,2,3,4
	CL = 19	CL = 22	tCK(AVG)	0.75	<0.833	1,2,3,4
	CL = 20	CL = 23	tCK(AVG)	0.75	<0.833	1,2,3
CWL = 16,20	CL = 20	CL = 24	tCK(AVG)	Reserved		1,2,3,4
	CL = 22	CL = 26	tCK(AVG)	0.625	<0.75	1,2,3,4
	CL = 24	CL = 28	tCK(AVG)	0.625	<0.75	1,2,3
Supported CL Settings			10,11,12,13,14,15,16,17,18,19,20,22,24		nCK	13
Supported CL Settings with Read DBI			12,13,14,15,16,18,19,20,21,22,23,24,26,28		nCK	
Supported CWL Settings			9,10,11,12,14,16,18,20		nCK	

Speed Bin Table Note

Absolute Specifications

- $VDDQ = VDD = 1.20V \pm 0.06 V$
 - $VPP = 2.5V$ (2.375V min, 2.75V max)
 - The values defined with above-mentioned table are DLL ON case.
 - DDR4- 2666 and 3200 Speed Bin Tables are valid only when Gear Down Mode is disabled.
1. The CL setting and CWL setting result in tCK(avg).MIN and tCK(avg).MAX requirements. When making a selection of tCK(avg), both need to be fulfilled: Requirements from CL setting as well as requirements from CWL setting.
 2. tCK(avg).MIN limits: Since CAS Latency is not purely analog - data and strobe output are synchronized by the DLL-all possible intermediate frequencies may not be guaranteed. CL in clock cycle is calculated from tAA following rounding algorithm defined in JEDEC-79C Section 13.5.
 3. tCK(avg).MAX limits: Calculate $tCK(avg) = tAA.MAX / CL \text{ SELECTED}$ and round the resulting tCK(avg) down to the next valid speed bin (i.e. 1.5ns or 1.25ns or 1.071 ns or 0.937 ns or 0.833 ns). This result is tCK(avg).MAX corresponding to CL SELECTED.
 4. 'Reserved' settings are not allowed. User must program a different value.
 5. 'Optional' settings allow certain devices in the industry to support this setting, however, it is not a mandatory feature. Refer to supplier's data sheet and/or the DIMM SPD information if and how this setting is supported.
 6. Any DDR4-1866 speed bin also supports functional operation at lower frequencies as shown in the corresponding table which are not subject to Production Tests but verified by Design/ Characterization.
 7. Any DDR4-2133 speed bin also supports functional operation at lower frequencies as shown in the corresponding table which are not subject to Production Tests but verified by Design/ Characterization.
 8. Any DDR4-2400 speed bin also supports functional operation at lower frequencies as shown in the corresponding table which are not subject to Production Tests but verified by Design/ Characterization.
 9. Any DDR4-2666 speed bin also supports functional operation at lower frequencies as shown in the corresponding table which are not subject to Production Tests but verified by Design/ Characterization.
 10. Any DDR4-3200 speed bin also supports functional operation at lower frequencies as shown in the corresponding table which are not subject to Production Tests but verified by Design/ Characterization.
 11. DDR4-1600 AC timing apply if DRAM operates at lower than 1600 MT/s data rate.
 12. Parameters apply from tCK(avg)min to tCK(avg)max at all standard JEDEC clock period values as stated in the Speed Bin Tables.
 13. CL number in parentheses, it means that these numbers are optional.
 14. DDR4 SDRAM supports CL=9 as long as a system meets tAA(min).
 15. Each speed bin lists the timing requirements that need to be supported in order for a given DRAM to be JEDEC compliant. JEDEC compliance does not require support for all speed bins within a given speed. JEDEC compliance requires meeting the parameters for a least one of the listed speed bins.

6.2 tREFI and tRFC Parameters

Table 6-2. tREFI and tRFC Parameters

Refresh Mode	Parameter		8Gb	Unit
	tREFI(base)		7.8	μs
1 x mode	tREFI1	0 °C ≤ TCASE ≤ 85 °C	tREFI(base)	μs
		85 °C < TCASE ≤ 95 °C	tREFI(base)/2	μs
		tRFC1(min)	350	ns
2 x mode	tREFI2	0 °C ≤ TCASE ≤ 85 °C	tREFI(base)/2	μs
		85 °C < TCASE ≤ 95 °C	tREFI(base)/4	μs
		tRFC2(min)	260	ns
4 x mode	tREFI4	0 °C ≤ TCASE ≤ 85 °C	tREFI(base)/4	μs
		85 °C < TCASE ≤ 95 °C	tREFI(base)/8	μs
		tRFC4(min)	160	ns

7 IDD AND IDDQ SPECIFICATION PARAMETERS AND TEST CONDITIONS

7.1 IDD, IPP and IDDQ Measurement Conditions

Figure 7-1 shows the setup and test load for IDD, IPP and IDDQ measurements.

- IDD currents (such as IDD0, IDD0A, IDD1, IDD1A, IDD2N, IDD2NA, IDD2NL, IDD2NT, IDD2P, IDD2Q, IDD3N, IDD3NA, IDD3P, IDD4R, IDD4RA, IDD4W, IDD4WA, IDD5B, IDD5F2, IDD5F4, IDD6N, IDD6E, IDD6R, IDD6A, IDD7 and IDD8) are measured as time-averaged currents with all VDD balls of the DDR4 SDRAM under test tied together. Any IPP or IDDQ current is not included in IDD currents.
- IPP currents have the same definition as IDD except that the current on the VPP supply is measured.
- IDDQ currents (such as IDDQ2NT and IDDQ4R) are measured as time-averaged currents with all VDDQ balls of the DDR4 SDRAM under test tied together. Any IDD current is not included in IDDQ currents.

Attention: IDDQ values cannot be directly used to calculate IO power of the DDR4 SDRAM. They can be used to support correlation of simulated IO power to actual IO power as outlined in Figure 7-2. In DRAM module application, IDDQ cannot be measured separately since VDD and VDDQ are using one merged-power layer in Module PCB.

For IDD, IPP and IDDQ measurements, the following definitions apply:

- “0” and “Low” is defined as $V_{IN} \leq V_{ILAC}(\max)$.
- “1” and “High” is defined as $V_{IN} \geq V_{IHAC}(\min)$.
- “MID-LEVEL” is defined as inputs are $V_{REF} = V_{DD} / 2$.
- Basic IDD, IPP and IDDQ Measurement Conditions are described in Table 9-2.
- Detailed IDD, IPP and IDDQ Measurement-Loop Patterns are described in Table 9-3.
- IDD Measurements are done after properly initializing the DDR4 SDRAM. This includes but is not

limited to setting

RON = RZQ/7 (34 Ohm in MR1);

RTT_NOM = RZQ/6 (40 Ohm in MR1);

RTT_WR = RZQ/2 (120 Ohm in MR2);

RTT_PARK = Disable;

Qoff = 0B (Output Buffer enabled) in MR1;

TDQS_t disabled in MR1;

CRC disabled in MR2;

CA parity feature disabled in MR5;

Gear down mode disabled in MR3;

Read/Write DBI disabled in MR5;

DM disabled in MR5

• Attention: The IDD, IPP and IDDQ Measurement-Loop Patterns need to be executed at least one time before actual IDD or IDDQ measurement is started.

• Define D = {CS_n, ACT_n, RAS_n, CAS_n, WE_n}: = {High, Low, Low, Low, Low}

• Define D# = {CS_n, ACT_n, RAS_n, CAS_n, WE_n}: = {High, High, High, High, High}; apply invert of BG/BA changes when directed above.

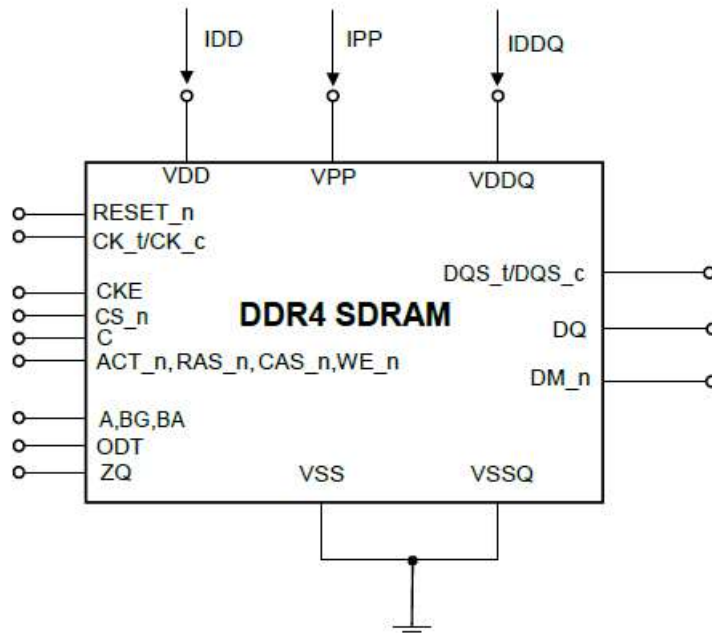


Figure 7-1. Measurement Setup and Test Load for IDD, IPP and IDDQ Measurements

Note:

DIMM level Output test load condition may be different from above

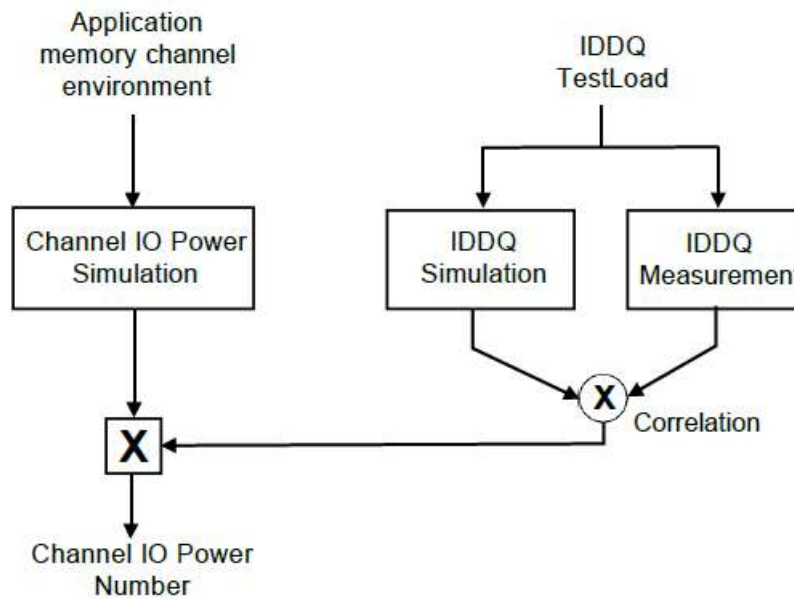


Figure 7-2. Correlation from Simulated Channel IO Power to Actual Channel IO Power Supported by IDDQ Measurement.

Table 7-1. Timings used for IDD, IPP and IDDQ Measurement-Loop Patterns

Symbol	DDR4-2666				DDR4-3200			Unit
	17-17-17	18-18-18	19-19-19	20-20-20	20-20-20	22-22-22	24-24-24	
tCK	0.75				0.625			ns
CL	17	18	19	20	20	22	24	nCK
CWL	16	16	18	18	18	20	20	nCK
nRCD	17	18	19	20	20	22	24	nCK
nRC	60	61	62	63	72	74	76	nCK
nRAS	43				52			nCK
nRP	17	18	19	20	20	22	24	nCK
nFAW	X4	16			16			nCK
	X8	28			34			nCK
	X16	40			48			nCK
nRRDS	X4	4			4			nCK
	X8	4			4			nCK
	X16	8			9			nCK
nRRDL	X4	7			8			nCK
	X8	7			8			nCK
	X16	9			11			nCK
tCCD S	4			4			nCK	
tCCD L	7			8			nCK	
tWTR S	4			4			nCK	
tWTR L	10			12			nCK	
nRFC 2Gb	21			256			nCK	
nRFC 4Gb	34			416			nCK	
nRFC 8Gb	46			560			nCK	
nRFC 16Gb	73			880			nCK	

Table 7-2. Basic IDD, IPP and IDDQ Measurement Conditions

Symbol	Description
DD0	<p>Operating One Bank Active-Precharge Current (AL=0) CKE: High; External clock: On; tCK, nRC, nRAS, nRCD, CL</p> <p>BL: 8¹; AL: 0; CS_n: High between ACT and PRE; Command, Address, Bank Group Address, Bank Address Inputs: partially toggling according to Table 7-3. Data IO: VDDQ; DM_n: stable at 1; Bank Activity: Cycling with one bank active at a time: 0,0,1,1,2,2,... (see Table 7-3); Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: stable at 0; Pattern Details: see Table 7-3.</p>
IDD0A	<p>Operating One Bank Active-Precharge Current (AL=CL-1) AL = CL-1, Other conditions: see IDD0</p>
IPP0	<p>Operating One Bank Active-Precharge IPP Current</p> <p>Same condition with IDD0</p>
IDD1	<p>Operating One Bank Active-Read-Precharge Current (AL=0) CKE: High; External clock: On; tCK, nRC, nRAS, nRCD, CL</p> <p>BL: 8¹; AL: 0; CS_n: High between ACT, RD and PRE; Command, Address, Bank Group Address, Bank Address Inputs, Data IO: partially toggling according to Table 7-4. DM_n: stable at 1; Bank Activity: Cycling with one bank active at a time: 0,0,1,1,2,2,... (see Table 7-4); Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: stable at 0; Pattern Details: see Table 7-4.</p>
IDD1A	<p>Operating One Bank Active-Read-Precharge Current (AL=CL-1) AL = CL-1, Other conditions: see IDD1</p>
IPP1	<p>Operating One Bank Active-Read-Precharge IPP Current</p> <p>Same condition with IDD1</p>

Symbol	Description
IDD2N	<p>Precharge Standby Current (AL=0) CKE: High; External clock: On; tCK, CL BL: 8¹; AL: 0; CS_n: stable at 1; Command, Address, Bank Group Address, Bank Address Inputs: partially toggling according to Table7-5. Data IO: VDDQ; DM_n: stable at 1; Bank Activity: all banks closed; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: stable at 0; Pattern Details: see Table 7-5</p>
IDD2NA	<p>Precharge Standby Current (AL=CL-1) AL = CL-1, Other conditions: see IDD2N</p>
IPP2N	<p>Precharge Standby IPP Current</p>
IDD2NT	<p>Precharge Standby ODT Current CKE: High; External clock: On; tCK, CL BL: 8¹; AL: 0; CS_n: stable at 1; Command, Address, Bank Group Address, Bank Address Inputs: partially toggling according to Table 7-6. Data IO: VSSQ; DM_n: stable at 1; Bank Activity: all banks closed; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal:ogging according to Table 7-6 Pattern Details: Table 7-6</p>
IDDQ2NT(Optional)	<p>Precharge Standby ODT IDDQ Current Same definition like for IDD2NT, however measuring IDDQ current instead of IDD current</p>
IDD2NL	<p>Precharge Standby Current with CAL enabled Same definition like for IDD2N, CAL enabled³</p>
IDD2NG	<p>Precharge Standby Current with Gear Down mode enabled Same definition like for IDD2N, Gear Down mode enabled^{3,5}</p>
IDD2ND	<p>Precharge Standby Current with DLL disabled Same definition like for IDD2N, DLL disabled³</p>
IDD2N_par	<p>Precharge Standby Current with CA parity enabled Same definition like for IDD2N, CA parity enabled³</p>

Symbol	Description
IDD2P	<p>Precharge Power-Down Current CKE: Low; External clock: On; tCK, CL BL: 8¹; AL: 0; CS_n: stable at 1; Command, Address, Bank Group Address, Bank Address Inputs: stable at 0; Data IO: VDDQ; DM_n: stable at 1;</p>
IPP2P	<p>Precharge Power-Down IPP Current Same condition with IDD2P</p>
IDD2Q	<p>Precharge Quiet Standby Current CKE: High; External clock: On; tCK, CL BL: 8¹; AL: 0; CS_n: stable at 1; Command, Address, Bank Group Address, Bank Address Inputs: stable at 0; Data IO: VDDQ; DM_n: stable at 1; Bank Activity: all banks closed; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: stable at 0</p>
IDD3N	<p>Active Standby Current CKE: High; External clock: On; tCK, CL BL: 8¹; AL: 0; CS_n: stable at 1; Command, Address, Bank Group Address, Bank Address Inputs: partially toggling according to Table 7-5. Data IO: VDDQ; DM_n: stable at 1; Bank Activity: all banks open; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: stable at 0; Pattern Details: see Table 7-5</p>
IDD3NA	<p>Active Standby Current (AL=CL-1) AL = CL-1, Other conditions: see IDD3N</p>
IPP3N	<p>Active Standby IPP Current Same condition with IDD3N</p>

Symbol	Description
IDD3P	<p>Active Power-Down Current CKE: Low; External clock: On; tCK, CL BL: 8¹; AL: 0; CS_n: stable at 1; Command, Address, Bank Group Address, Bank Address Inputs: stable at 0; Data IO: VDDQ; DM_n: stable at 1; Bank Activity: all banks open; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: stable at 0</p>
IPP3P	<p>Active Power-Down IPP Current Same condition with IDD3P</p>
IDD4R	<p>Operating Burst Read Current CKE: High; External clock: On; tCK, CL BL: 8²; AL: 0; CS_n: High between RD; Command, Address, Bank Group Address, Bank Address Inputs: partially toggling according to Table 7-7. Data IO: seamless Read data burst with different data between one burst and the next one according to Table 7-7. DM_n: stable at 1; Bank Activity: all banks open, RD commands cycling through banks: 0,0,1,1,2,2,... (see Table 7-7) Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: stable at 0; Pattern Details: see Table 7-7</p>
IDD4RA	<p>Operating Burst Read Current (AL=CL-1) AL = CL-1, Other conditions: see IDD4R</p>
IDD4RB	<p>Operating Burst Read Current with Read DBI Read DBI enabled³, Other conditions: see IDD4R</p>
IPP4R	<p>Operating Burst Read IPP Current Same condition with IDD4R</p>
IDDQ4R (Optional)	<p>Operating Burst Read IDDQ Current Same definition like for IDD4R, however measuring IDDQ current instead of IDD current</p>
IDDQ4RB (Optional)	<p>Operating Burst Read IDDQ Current with Read DBI Same definition like for IDD4RB, however measuring IDDQ current instead of IDD current</p>

Symbol	Description
IDD4W	<p>Operating Burst Write Current</p> <p>CKE: High;</p> <p>External clock: On;</p> <p>tCK, CL</p> <p>BL: 8¹;</p> <p>AL: 0;</p> <p>CS_n: High between WR;</p> <p>Command, Address, Bank Group Address, Bank Address Inputs: partially toggling according to Table 7-8;</p> <p>Data IO: seamless Write data burst with different data between one burst and the next one according to Table 7-8</p> <p>DM_n: stable at 1;</p> <p>Bank Activity: all banks open, WR commands cycling through banks: 0,0,1,1,2,2,... (Table 7-8);</p> <p>Output Buffer and RTT: Enabled in Mode Registers²;</p> <p>ODT Signal: stable at High;</p> <p>Pattern Details: see Table 7-8</p>
IDD4WA	Operating Burst Write Current (AL=CL-1) AL = CL-1, Other conditions: see IDD4W
IDD4WB	Operating Burst Write Current with Write DBI Write DBI enabled³, Other conditions: see IDD4W
IDD4WC	Operating Burst Write Current with Write CRC Write CRC enabled³, Other conditions: see IDD4W
IDD4W _{par}	Operating Burst Write Current with CA Parity CA Parity enabled³, Other conditions: see IDD4W
IPP4W	Operating Burst Write IPP Current Same condition with IDD4W
IDD5B	<p>Burst Refresh Current (1X REF)</p> <p>CKE: High;</p> <p>External clock: On;</p> <p>tCK, CL nRFC</p> <p>BL: 8¹;</p> <p>AL: 0;</p> <p>CS_n: High between REF;</p> <p>Command, Address, Bank Group Address, Bank Address Inputs: partially toggling according to Table 7-10.</p> <p>Data IO: VDDQ;</p> <p>DM_n: stable at 1;</p> <p>Bank Activity: REF command every nRFC (see Table 7-10);</p> <p>Output Buffer and RTT: Enabled in Mode Registers²;</p> <p>ODT Signal: stable at 0;</p> <p>Pattern Details: see Table 7-10</p>
IPP5B	Burst Refresh Write IPP Current (1X REF) Same condition with IDD5B
IDD5F2	Burst Refresh Current (2X REF) tRFC=tRFC_{x2}, Other conditions: see IDD5B
IPP5F2	Burst Refresh Write IPP Current (2X REF) Same condition with IDD5F2
IDD5F4	Burst Refresh Current (4X REF) tRFC=tRFC_{x4}, Other conditions: see IDD5B
IPP5F4	Burst Refresh Write IPP Current (4X REF) Same condition with IDD5F4

Symbol	Description
IDD6N	<p>Self Refresh Current: Normal Temperature Range <i>T</i>_{CASE}: 0 - 85°C; Low Power Array Self Refresh (LP ASR): Normal⁴; CKE: Low; External clock: Off; CK_t and CK_c#: Low; CL BL: 8¹; AL: 0; CS_n#, Command, Address, Bank Group Address, Bank Address, Data IO: High; DM_n: stable at 1; Bank Activity: Self-Refresh operation; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: MID-LEVEL</p>
IPP6N	<p>Self Refresh IPP Current: Normal Temperature Range Same condition with IDD6N</p>
IDD6E	<p>Self-Refresh Current: Extended Temperature Range) <i>T</i>_{CASE}: 0 - 95°C; Low Power Array Self Refresh (LP ASR): Extended⁴; CKE: Low; External clock: Off; CK_t and CK_c#: Low; CL BL: 8¹ AL: 0; CS_n, Command, Address, Bank Group Address, Bank Address, Data IO: High; DM_n: stable at 1; Bank Activity: Extended Temperature Self-Refresh operation; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: MID-LEVEL</p>
IPP6E	<p>Self Refresh IPP Current: Extended Temperature Range Same condition with IDD6E</p>
IDD6R	<p>Self-Refresh Current: Reduced Temperature Range <i>T</i>_{CASE}: 0 - TBD (~35-45) °C; Low Power Array Self Refresh (LP ASR): Reduced⁴; CKE: Low; External clock: Off; CK_t and CK_c#: Low; CL BL: 8¹ AL: 0; CS_n#, Command, Address, Bank Group Address, Bank Address, Data IO: High; DM_n: stable at 1; Bank Activity: Extended Temperature Self-Refresh operation; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: MID-LEVEL</p>
IPP6R	<p>Self Refresh IPP Current: Reduced Temperature Range Same condition with IDD6R</p>

Symbol	Description
IDD6A	<p>Auto Self-Refresh Current <i>T</i>_{CASE}: 0 - 95°C; Low Power Array Self Refresh (LPASR): Auto⁴; Partial Array Self-Refresh (PASR): Full Array; CKE: Low; External clock: Off; CK_t and CK_c#: Low; CL BL: 8¹ AL: 0; CS_n#, Command, Address, Bank Group Address, Bank Address, Data IO: High; DM_n: stable at 1; Bank Activity: Auto Self-Refresh operation; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: MID-LEVEL</p>
IPP6A	<p>Auto Self-Refresh IPP Current Same condition with IDD6A</p>
IDD7	<p>Operating Bank Interleave Read Current CKE: High; External clock: On; tCK, nRC, nRAS, nRCD, nRRD, nFAW, CL BL: 8¹; AL: CL-1; CS_n: High between ACT and RDA; Command, Address, Bank Group Address, Bank Address Inputs: partially toggling according to Table 7-11; Data IO: Read data bursts with different data between one burst and the next one according to Table 7-11; DM_n: stable at 1; Bank Activity: two times interleaved cycling through banks (0, 1, ...7) with different addressing, see Table 7-11; Output Buffer and RTT: Enabled in Mode Registers²; ODT Signal: stable at 0; Pattern Details: see Table 7-11.</p>
IPP7	<p>Operating Bank Interleave Read IPP Current Same condition with IDD7</p>
IDD8	<p>Maximum Power Down Current TBD</p>
IPP8	<p>Maximum Power Down IPP Current Same condition with IDD8</p>

7.1.1 IDD0, IDD0A and IPP0 Measurement-Loop Pattern

Table 7-3. IDD0, IDD0A and IPP0 Measurement-Loop Pattern¹

CK_t /CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴		
Toggling	Static High	0	0	ACT	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-	
			1	D, D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-
			3	D_#,D_#	1	1	1	1	1	1	0	0	3 ²	3	0	0	0	7	F	0	-	
			repeat pattern 1...4 unit nRAS -1, truncate if necessary																			
			nRAS	PRE	0	1	0	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0
		repeat pattern 1...4 unit nRC -1, truncate if necessary																				
		1	1*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 1, BA[1:0] = 1 instead																		
		2	2*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 2 instead																		
		3	3*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 1, BA[1:0] = 3 instead																		
		4	4*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 1 instead																		
		5	5*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 1, BA[1:0] = 2 instead																		
		6	6*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 3 instead																		
		7	7*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 1, BA[1:0] = 0 instead																		
		8	8*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 0 instead																		
		9	9*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 1 instead																		
10	10*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 2 instead																				
11	11*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 3 instead																				
12	12*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 1 instead																				
13	13*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 2 instead																				
14	14*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 3 instead																				
15	15*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 0 instead																				

For x4 and x8 only

- Note:
1. DQS_t, DQS_c are VDDQ.
 2. BG1 is Don't Care for x16 device
 3. C[2:0] are used only for 3DS device
 4. DQ signals are VDDQ

7.1.2 IDD1, IDD1A and IPP1 Measurement-Loop Pattern

Table 7-4. IDD1, IDD1A and IPP1 Measurement-Loop Pattern¹

CK_t / CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴			
Toggling	Static High	0	0	ACT	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-		
			1,2	D,D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-	
			3,4	D_#,D_#	1	1	1	1	1	1	1	0	0	3 ^b	3	0	0	0	7	F	0	-	
			repeat pattern 1...4 unit nRCD - AL - 1, truncate if necessary																				
			nRCD -AL	RD	0	1	1	0	1	0	1	0	0	0	0	0	0	0	0	0	0	0	D0=00, D1=FF D2=FF, D3=00 D4=FF, D5=00 D6=00, D7=FF
			repeat pattern 1...4 unit nRAS -1, truncate if necessary																				
			nRAS	PRE	0	1	0	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-
			repeat pattern nRC + 1...4 unit 1*nRC + nRAS -1, truncate if necessary																				
			1*nRC + 0	ACT	0	0	0	1	1	0	0	1	1	0	1	1	0	0	0	0	0	0	-
			1*nRC + 1,2	D,D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-
		1*nRC + 3,4	D_#,D_#	1	1	1	1	1	1	1	0	0	3 ^b	3	0	0	0	7	F	0	-		
		repeat pattern nRC + 1...4 unit 1*nRC + nRAS -1, truncate if necessary																					
		1*nRC + nRCD - AL	RD	0	1	1	0	1	0	1	0	0	1	1	0	0	0	0	0	0	0	D0=FF, D1=00 D2=00, D3=FF	
		repeat pattern 1...4 unit nRAS -1, truncate if necessary																					
		1*nRC + nRAS	PRE	0	1	0	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-	
		repeat nRC + 1...4 unit 2*nRC -1, truncate if necessary																					
		2	2*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 2 instead																			
		3	3*nRC	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 3 instead																			
		4	4*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 1 instead																			
		5	5*nRC	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 2 instead																			
6	6*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 3 instead																					
8	7*nRC	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 0 instead																					
9	9*nRC	repeat Sub-Loop 1, use BG[1:0] ² = 2, BA[1:0] = 0 instead																					

For x4 and x8 only

CK_t /CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴
		10	10*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 1 instead																
		11	11*nRC	repeat Sub-Loop 1, use BG[1:0] ² = 2, BA[1:0] = 2 instead																
		12	12*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 3 instead																
		13	13*nRC	repeat Sub-Loop 1, use BG[1:0] ² = 2, BA[1:0] = 1 instead																
		14	14*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 2 instead																
		15	15*nRC	repeat Sub-Loop 1, use BG[1:0] ² = 2, BA[1:0] = 3 instead																
		16	16*nRC	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 0 instead																

Note:

1. DQS_t, DQS_c are used according to RD Commands, otherwise VDDQ.
2. BG1 is Don't Care for x16 device
3. C[2:0] are used only for 3DS device
4. Burst Sequence driven on each DQ signal by Read Command. Outside burst operation, DQ signals are VDDQ.

7.1.3 IDD2N, IDD2NA, IDD2NL, IDD2NG, IDD2ND, IDD2N_par, IPP2, IDD3N, IDD3NA and IDD3P Measurement-Loop Pattern

Table 7-5. IDD2N, IDD2NA, IDD2NL, IDD2NG, IDD2ND, IDD2N_par, IPP2, IDD3N, IDD3NA and IDD3P Measurement-Loop Pattern¹

CK_t/CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴		
Toggling	Static High	0	0	D,D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	
			1	D,D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
			2	D_#D_#	1	1	1	1	1	1	0	0	0	3 ²	3	0	0	0	7	F	0	0
			3	D_#D_#	1	1	1	1	1	1	0	0	0	3 ²	3	0	0	0	7	F	0	0
		1	4-7	repeat Sub-Loop 0, use BG[1:0] ² = 1, BA[1:0] = 1 instead																		
		2	8-11	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 2 nstead																		
		3	12-15	repeat Sub-Loop 0, use BG[1:0] ² = 1, BA[1:0] = 3 instead																		
		4	16-19	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 1 instead																		
		5	20-23	repeat Sub-Loop 0, use BG[1:0] ² = 1, BA[1:0] = 2 instead																		
		6	24-27	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 3 instead																		
		7	28-31	repeat Sub-Loop 0, use BG[1:0] ² = 1, BA[1:0] = 0 instead																		
		8	32-35	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 0 instead																		
		9	36-39	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 1 instead																		
		10	40-43	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 2 instead																		
		11	44-47	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 3 instead																		
		12	48-51	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 1 instead																		
13	52-55	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 2 instead																				
14	56-59	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 3 instead																				
15	60-63	repeat Sub-Loop 0, use BG[1:0] ² = 3, BA[1:0] = 0 instead																				

Note:

1. DQS_t, DQS_c are VDDQ.
2. BG1 is Don't Care for x16 device
3. C[2:0] are used only for 3DS device
4. DQ signals are VDDQ.

7.1.4 IDD2NT and IDDQ2NT Measurement-Loop Pattern

Table 7-6. IDD2NT and IDDQ2NT Measurement-Loop Pattern¹

CK_t/CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴		
Toggling	Static High	0	0	D,D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-	
			1	D,D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-
			2	D_#,D_#	1	1	1	1	1	1	1	0	0	3 ²	3	0	0	0	7	F	0	-
			3	D_#,D_#	1	1	1	1	1	1	1	0	0	3 ²	3	0	0	0	7	F	0	-
		1	4-7	repeat Sub-Loop 0, but ODT = 1 and BG[1:0] ² = 1, BA[1:0] = 1 instead																		
		2	8-11	repeat Sub-Loop 0, but ODT = 0 and BG[1:0] ² = 0, BA[1:0] = 2 instead																		
		3	12-15	repeat Sub-Loop 0, but ODT = 1 and BG[1:0] ² = 1, BA[1:0] = 3 instead																		
		4	16-19	repeat Sub-Loop 0, but ODT = 0 and BG[1:0] ² = 0, BA[1:0] = 1 instead																		
		5	20-23	repeat Sub-Loop 0, but ODT = 1 and BG[1:0] ² = 1, BA[1:0] = 2 instead																		
		6	24-27	repeat Sub-Loop 0, but ODT = 0 and BG[1:0] ² = 0, BA[1:0] = 3 instead																		
		7	28-31	repeat Sub-Loop 0, but ODT = 1 and BG[1:0] ² = 1, BA[1:0] = 0 instead																		
		8	32-35	repeat Sub-Loop 0, but ODT = 0 and BG[1:0] ² = 2, BA[1:0] = 0 instead																		
		9	36-39	repeat Sub-Loop 0, but ODT = 1 and BG[1:0] ² = 3, BA[1:0] = 1 instead																		
		10	40-43	repeat Sub-Loop 0, but ODT = 0 and BG[1:0] ² = 2, BA[1:0] = 2 instead																		
		11	44-47	repeat Sub-Loop 0, but ODT = 1 and BG[1:0] ² = 3, BA[1:0] = 3 instead																		
		12	48-51	repeat Sub-Loop 0, but ODT = 0 and BG[1:0] ² = 2, BA[1:0] = 1 instead																		
13	52-55	repeat Sub-Loop 0, but ODT = 1 and BG[1:0] ² = 3, BA[1:0] = 2 instead																				
14	56-59	repeat Sub-Loop 0, but ODT = 0 and BG[1:0] ² = 2, BA[1:0] = 3 instead																				
15	60-63	repeat Sub-Loop 0, but ODT = 1 and BG[1:0] ² = 3, BA[1:0] = 0 instead																				

Note:

1. DQS_t, DQS_c are VDDQ.
2. BG1 is Don't Care for x16 device
3. C[2:0] are used only for 3DS device
4. DQ signals are VDDQ.

7.1.5 IDD4R, IDDR4RA, IDD4RB and IDDQ4R Measurement-Loop Pattern¹

Table7-7. IDD4R, IDDR4RA, IDDR4RB and IDDQ4R Measurement-Loop Pattern¹

CK_t/CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ¹	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴			
Toggling	Static High	0	0	RD	0	1	1	0	1	0	0	0	0	0	0	0	0	0	0	0	D0=00, D1=FF D2=FF, D3=00 D4=FF, D5=00 D6=00, D7=FF		
			1	D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-	
			2,3	D_#D_#	1	1	1	1	1	1	0	0	3 ²	3	0	0	0	0	7	F	0	-	
		1	4	RD	0	1	1	0	1	0	0	0	1	1	0	0	0	0	7	F	0	D0=FF, D1=00 D2=00, D3=FF D4=00, D5=FF D6=FF, D7=00	
			5	D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-
			6,7	D_#D_#	1	1	1	1	1	1	0	0	3	3	0	0	0	0	7	F	0	-	
		2	8-11	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 2 instead																			
		3	12-15	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 3 instead																			
		4	16-19	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 1 instead																			
		5	20-23	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 2 instead																			
		6	24-27	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 3 instead																			
		7	28-31	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 0 instead																			
		8	32-35	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 0 instead																			
		9	36-39	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 1 instead																			
		10	40-43	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 2 instead																			
11	44-47	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 3 instead																					
12	48-51	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 1 instead																					
13	52-55	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 2 instead																					
14	56-59	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 3 instead																					
15	60-63	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 0 instead																					

Note:

1. DQS_t, DQS_c are used according to RD Commands, otherwise VDDQ.
2. BG1 is Don't Care for x16 device
3. C[2:0] are used only for 3DS device
4. Burst Sequence driven on each OQ signal by Read Command

7.1.6 IDD4W, IDD4WA, IDD4WB and IDD4W_par Measurement-Loop Pattern

Table 7-8. IDD4W, IDDR4WA, IDDR4WB and IDD4W_par Measurement-Loop Pattern¹

CK_t /CK_c	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴		
Toggling	Static High	0	0	WR	0	1	1	0	1	0	0	0	0	0	0	0	0	0	0	D0=00, D1=FF D2=FF, D3=00 D4=FF, D5=00 D6=00, D7=FF	
			1	D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0		
			2,3	D_#,D_#	1	1	1	1	1	1	0	0	3 ²	3	0	0	0	7	F	0	
		1	4	WR	0	1	1	0	0	1	0	0	1	1	0	0	0	7	F	0	D0=FF, D1=00 D2=00, D3=FF
			5	D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	
			6,7	D_#,D_#	1	1	1	1	1	1	0	0	3 ²	3	0	0	0	7	F	0	
		2	8-11	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 2 instead																	
		3	12-15	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 3 instead																	
		4	16-19	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 1 instead																	
		5	20-23	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 2 instead																	
		6	24-27	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 3 instead																	
		7	28-31	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 0 instead																	
		8	32-35	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 0 instead																	
		9	36-39	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 1 instead																	
		10	40-43	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 2 instead																	
11	44-47	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 3 instead																			
12	48-51	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 1 instead																			
13	52-55	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 2 instead																			
14	56-59	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 3 instead																			
15	60-63	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 0 instead																			

Note:

1. DQS_t, DQS_c are used according to WR Commands, otherwise VDDQ.
2. BG1 is Don't Care for x16 device
3. C[2:0] are used only for 3DS device
4. Burst Sequence driven on each DQ signal by Write Command

7.1.7 IDD4WC Measurement-Loop Pattern

Table 7-9. IDD4WC Measurement-Loop Pattern¹

CK_t/CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴		
Toggling	Static High	0	0	WR	0	1	1	0	0	1	0	0	0	0	0	0	0	0	0	0	D0=00, D1=FF D2=FF, D3=00 D4=FF, D5=00 D6=00, D7=FF D8=CRC	
			1,2	D,D	1	0	0	0	0	0	1	0	0	0	0	0	0	0	0	0	0	-
			3,4	D_#,D_#	1	1	1	1	1	1	1	0	3 ²	3	3	0	0	0	7	F	0	-
			5	WR	0	1	1	0	0	0	1	0	1	1	1	0	0	0	7	F	0	D0=FF, D1=00 D2=00, D3=FF D4=00, D5=FF D6=FF, D7=00 D8=CRC
			6,7	D,D	1	0	0	0	0	0	1	0	0	0	0	0	0	0	0	0	0	-
			8,9	D_#,D_#	1	1	1	1	1	1	1	0	3	3	3	0	0	0	7	F	0	-
		2	10-14	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 2 instead																		
		3	15-19	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 3 instead																		
		4	20-24	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 1 instead																		
		5	25-29	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 2 instead																		
		6	30-34	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 3 instead																		
		7	35-39	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 0 instead																		
		8	40-44	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 0 instead																		
		9	45-49	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 1 instead																		
		10	50-54	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 2 instead																		
11	55-59	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 3 instead																				
12	60-64	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 1 instead																				
13	65-69	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 2 instead																				
14	70-74	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 3 instead																				
15	75-79	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 0 instead																				

Note:

1. DQS_t, DQS_c are VDDQ.
2. BG1 is Don't Care for x16 device
3. C[2:0] are used only for 3DS device
4. Burst Sequence driven on each DQ signal by Write Command.

7.1.8 IDD5B Measurement-Loop Pattern

Table 7-10. IDD5B Measurement-Loop Pattern¹

CK_t/CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴			
Toggling	Static High	0	0	REF	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-		
			1	D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-	
				1	2	D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-
					3	D_#,D_#	1	1	1	1	1	0	0	3 ²	3	0	0	0	7	F	0	0	-
					4	D_#,D_#	1	1	1	1	1	0	0	3 ²	3	0	0	0	7	F	0	0	-
					4-7	repeat pattern 1...4, use BG[1:0] ² = 1, BA[1:0] = 1 instead.																	
					8-11	repeat pattern 1...4, use BG[1:0] ² = 0, BA[1:0] = 2 instead.																	
					12-15	repeat pattern 1...4, use BG[1:0] ² = 1, BA[1:0] = 3 instead.																	
					16-19	repeat pattern 1...4, use BG[1:0] ² = 0, BA[1:0] = 1 instead.																	
					20-23	repeat pattern 1...4, use BG[1:0] ² = 1, BA[1:0] = 2 instead.																	
					24-27	repeat pattern 1...4, use BG[1:0] ² = 0, BA[1:0] = 3 instead.																	
					28-31	repeat pattern 1...4, use BG[1:0] ² = 1, BA[1:0] = 0 instead.																	
					32-35	repeat pattern 1...4, use BG[1:0] ² = 2, BA[1:0] = 0 instead.																	
					36-39	repeat pattern 1...4, use BG[1:0] ² = 3, BA[1:0] = 1 instead.																	
					40-43	repeat pattern 1...4, use BG[1:0] ² = 2, BA[1:0] = 2 instead.																	
					44-47	repeat pattern 1...4, use BG[1:0] ² = 3, BA[1:0] = 3 instead.																	
					48-51	repeat pattern 1...4, use BG[1:0] ² = 2, BA[1:0] = 1 instead.																	
					52-55	repeat pattern 1...4, use BG[1:0] ² = 3, BA[1:0] = 2 instead.																	
					56-59	repeat pattern 1...4, use BG[1:0] ² = 2, BA[1:0] = 3 instead.																	
					60-63	repeat pattern 1...4, use BG[1:0] ² = 3, BA[1:0] = 0 instead.																	
		2	64... nRFC -1	repeat Sub-Loop 1, Truncate, if necessary																			

Note:

1. DQS_t, DQS_c are VDDQ.
2. BG1 is Don't Care for x16 device
3. C[2:0] are used only for 3DS device
4. DQ signals are VDDQ

7.1.9 IDD7 Measurement-Loop Pattern

Table 7-11. IDD7 Measurement-Loop Pattern¹

CK_t/CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴				
Toggling	Static High	0	0	ACT	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	-			
			1	RDA	0	1	1	0	1	0	0	0	0	0	0	0	0	1	0	0	0	D0=00, D1=FF D2=FF, D3=00 D4=FF, D5=00 D6=00, D7=FF		
			2	D	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0			
			3	D_#	1	1	1	1	1	1	0	0	0	3 ²	3	0	0	0	7	F	0			
			...	repeat pattern 2...3, until nRRD - 1, if nRRD > 4. Truncate if necessary.																				
		1	nRRD	ACT	0	0	0	0	0	0	0	0	0	1	1	0	0	0	0	0	0	0		
			nRRD + 1	RDA	0	1	1	0	1	0	0	0	0	1	1	0	0	1	0	0	0	0	D0=FF, D1=00 D2=00, D3=FF D4=00, D5=FF D6=FF, D7=00	
			...	repeat pattern 2...3, until 2*nRRD - 1, if nRRD > 4. Truncate if necessary.																				
			2	2*nRRD	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 2 instead.																			
			3	3*nRRD	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 3 instead.																			
			4	4*nRRD	repeat pattern 2...3, until nFAW - 1, if nFAW > 4*nRRD. Truncate if necessary.																			
			5	nFAW	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 1 instead.																			
			6	nFAW + nRRD	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 2 instead.																			
			7	nFAW + 2*nRRD	repeat Sub-Loop 0, use BG[1:0] ² = 0, BA[1:0] = 3 instead.																			
			8	nFAW + 3*nRRD	repeat Sub-Loop 1, use BG[1:0] ² = 1, BA[1:0] = 0 instead.																			
9	nFAW + 4*nRRD	repeat Sub-Loop 4																						
10	2*nFAW	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 1 instead.																						
11	2*nFAW + nRRD	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 2 instead.																						
12	2*nFAW + 2*nRRD	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 3 instead.																						
13	2*nFAW + 3*nRRD	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 0 instead.																						
14	2*nFAW + 4*nRRD	repeat Sub-Loop 4																						
15	3*nFAW	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 1 instead.																						

For x4 and x8 only

CK_t /CK_c	CKE	Sub-Loop	Cycle Number	Command	CS_n	ACT_n	RAS_n/A16	CAS_n/A15	WE_n/A14	ODT	C[2:0] ³	BG[1:0] ²	BA[1:0]	A12/BC_n	A[17,13,11]	A[10]/AP	A[9:7]	A[6:3]	A[2:0]	Data ⁴
		1	3*nFAW +	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 2 instead.																
		1	3*nFAW +	repeat Sub-Loop 0, use BG[1:0] ² = 2, BA[1:0] = 3 instead.																
		1	3*nFAW +	repeat Sub-Loop 1, use BG[1:0] ² = 3, BA[1:0] = 0 instead.																
		1	3*nFAW +	repeat Sub-Loop 4																
		2	4*nFA	repeat pattern 2...3, until nRC - 1, if nRC > 4*nFAW. Truncate if necessary.																

Note:

1. DQS_t, DQS_c are VDDQ.
2. BG1 is Don't Care for x16 device
3. C[2:0] are used only for 3DS device
4. Burst Sequence driven on each DQ signal by Read Command. Outside burst operation, DQ signals are VDDQ.

7.2 IDD Specifications

IDD and IPP values are for full operation range of voltage and temperature unless otherwise noted.

Table 7-12. IDD and IDDQ Specifications (Rev. B)

Symbol	Width	DDR4-3200	Unit
IDD0	x16	TBD	mA
IDD0A	x16	TBD	mA
IDD1	x16	TBD	mA
IDD1A	x16	TBD	mA
IDD2N	x16	TBD	mA
IDD2NA	x16	TBD	mA
IDD2NT	x16	TBD	mA
IDD2NL	x16	TBD	mA
IDD2NG	x16	TBD	mA
IDD2ND	x16	TBD	mA
IDD2N_par	x16	TBD	mA
IDD2P	x16	TBD	mA
IDD2Q	x16	TBD	mA
IDD3N	x16	TBD	mA
IDD3NA	x16	TBD	mA
IDD3P	x16	TBD	mA
IDD4R	x16	TBD	mA
IDD4RA	x16	TBD	mA
IDD4RB	x16	TBD	mA
IDD4W	x16	TBD	mA
IDD4WA	x16	TBD	mA
IDD4WB	x16	TBD	mA
IDD4WC	x16	TBD	mA
IDD4W_par	x16	TBD	mA
IDD5B	x16	TBD	mA
IDD5F2	x16	TBD	mA
IDD5F4	x16	TBD	mA
IDD6N	x16	TBD	mA
IDD6E	x16	TBD	mA
IDD6R	x16	TBD	mA
IDD6A	x16	TBD	mA
IDD7	x16	TBD	mA
IDD8	x16	TBD	mA

Note:

1. Maximum limits are specified.
2. Users should refer to the DRAM supplier data sheet and/or the DIMM SPD to determine if DDR4 SDRAM devices support the following options or requirements referred to in this material.

Table 7-13. IPP Specifications (Rev. B)

Symbol	Width	DDR4-3200	Unit
IPP0	x16	TBD	mA
IPP1	x16	TBD	mA
IPP2N	x16	TBD	mA
IPP2P	x16	TBD	mA
IPP3N	x16	TBD	mA
IPP3P	x16	TBD	mA
IPP4R	x16	TBD	mA
IPP4W	x16	TBD	mA
IPP5B	x16	TBD	mA
IPP5F2	x16	TBD	mA
IPP5F4	x16	TBD	mA
IPP6N	x16	TBD	mA
IPP6E	x16	TBD	mA
IPP6R	x16	TBD	mA
IPP6A	x16	TBD	mA
IPP7	x16	TBD	mA
IPP8	x16	TBD	mA

Note:

1. Maximum limits are specified
2. Users should refer to the DRAM supplier data sheet and/or the DIMM SPD to determine if DDR4 SDRAM devices support the following options or requirements referred to in this material

7.2.1 IDD6 Specifications

Symbol	Temperature Range	Width	DDR4-3200	Unit	Note
IDD6R	0-45 °C	x16	TBD	mA	4,6,9
IDD6N	0-85 °C	x16	TBD	mA	3,4
IDD6E	0-95 °C	x16	TBD	mA	4,5,6
IDD6A	0-95 °C	x16	TBD	mA	4,6,7,8

Note:

1. Some IDD currents are higher for x16 organization due to larger page-size architecture.
2. Max. values for IDD currents considering worst case conditions of process, temperature and voltage.
3. Applicable for MR2 settings A6=0 and A7=0.
4. Supplier data sheets include a max value for IDD6.
5. Applicable for MR2 settings A6=0 and A7=1. IDD6E is only specified for devices which support the Extended Temperature Range feature.
6. Refer to the supplier data sheet for the value specification method (e.g. max, typical) for IDD6E and IDD6A
7. Applicable for MR2 settings A6=1 and A7=0. IDD6A is only specified for devices which support the Auto Self Refresh feature.
8. The number of discrete temperature ranges supported and the associated Ta – Tz values are supplier/design specific. Temperature ranges are specified for all supported values of TOPER. Refer to supplier data sheet for more information.
9. Applicable for MR2 settings MR2[A7:A6 = 01]: Reduced Temperature range. IDD6R is verified by design and characterization, and may not be subject to production test.

8 INPUT/OUTPUT CAPACITANCE

Table 8-1. Silicon Pad I/O Capacitance

Symbol	Parameter	3200		Unit	Note
		Min	Max		
C _{IO}	Input/output capacitance	0.55	1.00	pF	1,2,3
C _{DIO}	Input/output capacitance delta	-0.1	0.1	pF	1,2,3,11
C _{DDQS}	Input/output capacitance delta DQS _t and DQS _c	-	0.05	pF	1,2,3,5
C _{CK}	Input capacitance, CK _t and CK _c	0.2	0.55	pF	1,3
C _{DCK}	Input capacitance delta CK _t and CK _c	-	0.05	pF	1,3,4
C _I	Input capacitance (CTRL, ADD, CMD pins only)	0.2	0.55	pF	1,3,6
C _{DI_CTRL}	Input capacitance delta (All CTRL pins only)	-0.1	0.1	pF	1,3,7,8
C _{DI_ADD_CMD}	Input capacitance delta (All ADD/ CMD pins only)	-0.1	0.1	pF	1,2,9,10
C _{ALERT}	Input/output capacitance of ALERT	0.5	1.5	pF	1,3
C _{ZQ}	Input/output capacitance of ZQ	0.5	2.3	pF	1,3,12
C _{TEN}	Input capacitance OF TEN	0.2	2.3	pF	1,3,13

Note:

1. This parameter is not subject to production test. It is verified by design and characterization. The silicon only capacitance is validated by deembedding the package L & C parasitic. The capacitance is measured with VDD, VDDQ, VSS, VSSQ applied with all other signal pins floating. Measurement procedure TBD.
2. DQ, DM_n, DQS_T, DQS_C, TDQS_T, TDQS_C. Although the DM, TDQS_T and TDQS_C pins have different functions, the loading matches DQ and DQS.
3. This parameter applies to monolithic devices only; stacked/dual-die devices are not covered here.
4. Absolute value CK_T - CK_C.
5. Absolute value of C_{IO}(DQS_T) - C_{IO}(DQS_C).
6. C_I applies to ODT, CS_n, CKE, A0 - A17, BA0 - BA1, BG0 - BG1, RAS_n / A16, CAS_n / A15, WE_n / A14, ACT_n and PAR.
7. C_{DI_CTRL} applies to ODT, CS_n and CKE.
8. $C_{DI_CTRL} = C_i(CTRL) - 0.5 * (C_i(CLK_T) + C_i(CLK_C))$.
9. C_{DI_ADD_CMD} applies to, A0 - A17, BA0 - BA1, BG0 - BG1, RAS_n / A16, CAS_n / A15, WE_n / A14, ACT_n and PAR.
10. $C_{DI_ADD_CMD} = C_i(ADD_CMD) - 0.5 * (C_i(CLK_T) + C_i(CLK_C))$.
11. $C_{DIO} = C_{IO}(DQ, DM) - 0.5 * (C_{IO}(DQS_T) + C_{IO}(DQS_C))$.
12. Maximum external load capacitance on ZQ pin: tbd pF.
13. TEN pin may be DRAM internally pulled low through a weak pull-down resistor to VSS. In this case CTEN might not be valid and system shall verify TEN signal with Vendor specific information.

Table 8-2. DRAM Package Electrical Specifications (x16)

Symbol	Parameter	2666/3200		Unit	Note
		Min	Max		
Z _{IO}	Input/output Zpkg	45	85	Ω	1
T _{DIO}	Input/output Pkg Delay	14	45	ps	1
L _{IO}	Input/output Lpkg	-	3.4	nH	1,2
C _{IO}	Input/output Cpkg	-	0.82	pF	1,3
Z _{IO DQS}	DQS_t and DQS_c Zpkg	45	85	Ω	1
T _{DIO DQS}	DQS_t and DQS_c Pkg Delay	14	45	ps	1
L _{IO DQS}	DQS Lpkg	-	3.4	nH	1,2
C _{IO DQS}	DQS Cpkg	-	0.82	pF	1,3
DZ _{DIO DQS}	Delta Zpkg DQSU_t, DQSU_c	-	10	Ω	-
	Delta Zpkg DQSL_t, DQSU_c	-	10	Ω	-
DT _{DIO DQS}	Delta Delay DQSU_t, DQSU_c	-	5	ps	-
	Delta Delay DQSL_t, DQSU_c	-	5	ps	-
Z _{I CTRL}	Input-CTRL pins Zpkg	50	90	Ω	1
T _{dIADD_CMD}	Input-CTRL pins Pkg Delay	14	42	ps	1
L _{I CTRL}	Input CTRL Lpkg	-	3.4	nH	1,2
C _{I CTRL}	Input CTRL Cpkg	-	0.7	pF	1,3
Z _{IADD_CMD}	Input-CMD ADD pins Zpkg	50	90	Ω	1
T _{dIADD_CMD}	Input-CMD ADD pins Pkg Delay	14	52	ps	1
L _{IADD_CMD}	Input-CMD ADD Lpkg	-	3.9	nH	1,2
C _{IADD_CMD}	Input-CMD ADD Cpkg	-	0.86	pF	1,3

Note:

- 1 This parameter is not subject to production test. It is verified by design and characterization. The package parasitic(L & C) are validated using package only samples. The capacitance is measured with VDD, VDDQ, VSS, VSSQ shorted with all other signal pins floating. The inductance is measured with VDD, VDDQ, VSS and VSSQ shorted and all other signal pins shorted at the die side(not pin). Measurement procedure tbd
- 2 Package only impedance (Zpkg) is calculated based on the Lpkg and Cpkg total for a given pin where:

$$Z_{pkg} \text{ (total per pin)} = \sqrt{L_{pkg}/C_{pkg}}$$

- 3 Package only delay(Tpkg) is calculated based on Lpkg and Cpkg total for a given pin where:

$$T_{dpgk} \text{ (total per pin)} = \sqrt{L_{pkg}/C_{pkg}}$$

4 Z & Td IO applies to DQ, DM, DQS_C, DQS_T, TDQS_T and TDQS_C

5 This parameter applies to monolithic devices only; stacked/dual-die devices are not covered here

6 Absolute value of ZCK_t-ZCK_c for impedance(Z) or absolute value of TdCK_t-TdCK_c for delay(Td).

7 Absolute value of ZIO(DQS_t)-ZIO(DQS_c) for impedance(Z) or absolute value of TdIO(DQS_t)-TdIO(DQS_c) for delay(Td)

8 ZI & Td ADD CMD applies to A0-A17, BA0-BA1, BG0-BG1, RAS_n CAS_n, WE_n.

9 ZI & Td CTRL applies to ODT, CS_n and CKE.

10 This table applies to monolithic X4 and X8 devices.

- 11 Package implementations shall meet spec if the Zpkg and Pkg Delay fall within the ranges shown, and the maximum Lpkg and Cpkg do not exceed the maximum values shown.
- 12 It is assumed that Lpkg can be approximated as Lpkg = Zo*Td.
- 13 It is assumed that Cpkg can be approximated as Cpkg = Td/Zo.

9 ELECTRICAL CHARACTERISTICS & AC TIMING

9.1 Reference Load for AC Timing and Output Slew Rate

Figure 11-1 represents the effective reference load of 50 ohms used in defining the relevant AC timing parameters of the device as well as output slew rate measurements.

RON nominal of DQ, DQS_t and DQS_c drivers uses 34 ohms to specify the relevant AC timing parameter values of the device.

- The maximum DC High level of Output signal = 1.0*VDDQ
- The minimum DC Low level of Output signal = $\{34 / (34 + 50)\} * VDDQ = 0.4 * VDDQ$
- The nominal reference level of an Output signal can be approximated by the following:
- The center of maximum DC High and minimum DC Low = $\{(1 + 0.4) / 2\} * VDDQ = 0.7 * VDDQ$

The actual reference level of Output signal might vary with driver RON and reference load tolerances. Thus, the actual reference level or midpoint of an output signal is at the widest part of the output signal's eye. Prior to measuring AC parameters, the reference level of the verification tool should be set to an appropriate level.

It is not intended as a precise representation of any particular system environment or a depiction of the actual load presented by a production tester. System designers should use IBIS or other simulation tools to correlate the timing reference load to a system environment. Manufacturers correlate to their production test conditions, generally one or more coaxial transmission lines terminated at the tester electronics.

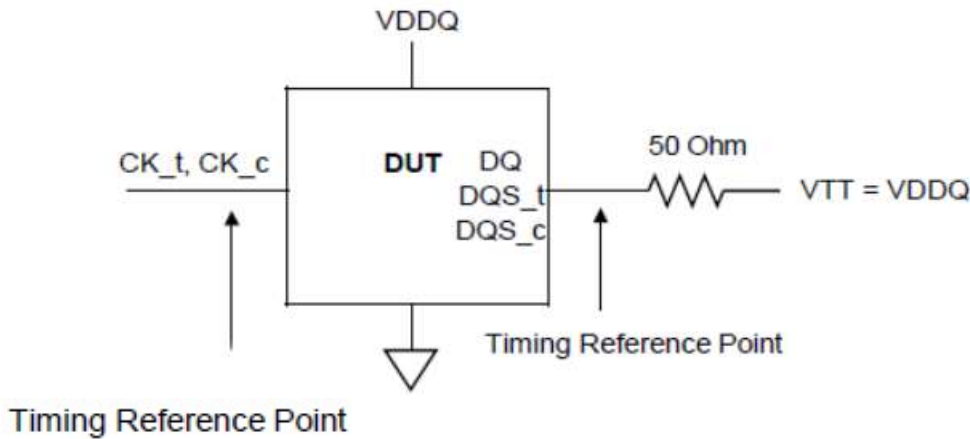


Figure 9-1. Reference Load for AC Timing and Output Slew Rate

9.2 tREFI

Average periodic Refresh interval (tREFI) of DDR4 SDRAM is defined as shown in Table 9-1 below.

Table 9-1. tREFI by Device Density

Parameter	Symbol	8Gb	Unit
Average periodic refresh interval	tREFI	0 °C ≤ TCASE ≤ 85 °C	7.8
		85 °C ≤ TCASE ≤ 95 °C	3.9

9.3 Clock Specification

The jitter specified is a random jitter meeting a Gaussian distribution. Input clocks violating the min/max values may result in malfunction of the DDR4 SDRAM device.

9.3.1 Definition for tCK(abs)

tCK(abs) is defined as the absolute clock period, as measured from one rising edge to the next consecutive rising edge. tCK(abs) is not subject to production test.

9.3.2 Definition for tCK(avg)

tCK(avg) is calculated as the average clock period across any consecutive 200 cycle window, where each clock period is calculated from rising edge to rising edge.

$$tCK(avg) = \left(\sum_{j=1}^N tCK(abs)j \right) / N \quad N=200$$

9.3.3 Definition for tCH(avg) and Tcl(avg)

tCH(avg) is defined as the average high pulse width, as calculated across any consecutive 200 high pulses.

$$tCH(avg) = \left(\sum_{j=1}^N tCHj \right) / (N \times tCK(avg)) \quad N=200$$

tCL(avg) is defined as the average low pulse width, as calculated across any consecutive 200 low pulses.

$$tCL(avg) = \left(\sum_{j=1}^N tCLj \right) / (N \times tCK(avg)) \quad N=200$$

9.3.4 Definition for tERR(nper)

tERR is defined as the cumulative error across n consecutive cycles of n x tCK(avg). tERR is not subject to production test.

9.4 Timing Parameters by Speed Grade

9.4.1 Timing Parameters by Speed Bin for DDR4-2666 to 3200

Table 9-2. Timing Parameters by Speed Bin for DDR4_2666 to 3200

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
Clock Timing									
Minimum Clock Cycle Time (DLL off mode)	tCK (DLL_OFF)	8	20	8	20	8	20	ns	
Average Clock Period	tCK (avg)	0.75	<0.800	0.682	<0.750	0.625	<0.682	n	35,36
Average high pulse width	tCH (avg)	0.48	0.52	0.48	0.52	0.48	0.52	tCK (avg)	
Average low pulse width	tCL (avg)	0.48	0.52	0.48	0.52	0.48	0.52	tCK (avg)	
Absolute Clock Period	tCK (abs)	Min = tCK(avg)min + tJIT(per)min_tot						tCK (avg)	
		Max = tCK(avg)max + tJIT(per)max_tot							
Absolute clock HIGH pulse width	tCH (abs)	0.45	-	0.45	-	0.45	-	tCK (avg)	23
Absolute clock LOW pulse width	tCL (abs)	0.45	-	0.45	-	0.45	-	tCK (avg)	24
Clock Period Jitter- total	JIT (per)_tot	-38	38	-34	34	-32	32	ps	25
Clock Period Jitter- deterministic	JIT (per)_dj	-19	19	-17	17	-16	16	ps	26
Clock Period Jitter during DLL locking period	tJIT (per, lck)	-30	30	-27	27	-25	25	ps	
Cycle to Cycle Period Jitter	tJIT (cc)	-	75	-	68	-	62	ps	25
Cycle to Cycle Period Jitter during DLL locking	tJIT (cc, lck)	-	60	-	55	-	50	ps	
Cumulative error across 2 cycles	tERR (2per)	-55	55	-50	50	-46	46	ps	
Cumulative error across 3 cycles	tERR (3per)	-66	66	-60	60	-55	55	ps	
Cumulative error across 4 cycles	tERR (4per)	-73	73	-66	66	-61	61	ps	
Cumulative error across 5 cycles	tERR (5per)	-78	78	-71	71	-65	65	ps	
Cumulative error across 6 cycles	tERR (6per)	-83	83	-75	75	-69	69	ps	
Cumulative error across 7 cycles	tERR (7per)	-87	87	-79	79	-73	73	ps	
Cumulative error across 8 cycles	tERR (8per)	-91	91	-83	83	-76	76	ps	
Cumulative error across 9 cycles	tERR (9per)	-94	94	-85	85	-78	78	ps	

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
Cumulative error across 10 cycles	tERR (10per)	-96	96	-88	88	-80	80	ps	
Cumulative error across 11 cycles	tERR (11per)	-99	99	-90	90	-83	83	ps	
Cumulative error across 12 cycles	tERR (12per)	-101	101	-92	92	-84	84	ps	
Cumulative error across 13 cycles	tERR (13per)	-103	103	-93	93	-86	86	ps	
Cumulative error across 14 cycles	tERR (14per)	-104	104	-95	95	-87	87	ps	
Cumulative error across 15 cycles	tERR (15per)	-106	106	-97	97	-89	89	ps	
Cumulative error across 16 cycles	tERR (16per)	-108	108	-98	98	-90	90	ps	
Cumulative error across 17 cycles	tERR (17per)	-110	110	-100	100	-92	92	ps	
Cumulative error across 18 cycles	tERR (18per)	-112	112	-101	101	-93	93	ps	
Cumulative error across n = 13, 14 . . . 49, 50 cycles	tERR (nper)	tERR (nper)min = ((1 + 0.68ln(n)) *tJIT (per)_total min)						ps	
		tERR (nper)max = ((1 + 0.68ln(n)) *tJIT (per)_total max)							
Command and Address setup time to CK_t, CK_c referenced to Vih(ac) / Vil(ac) levels	tIS (base)	55	-	48	-	40	-	ps	
Command and Address setup time to CK_t, CK_c referenced to Vref levels	tIS (Vref)	145	-	138	-	130	-	ps	
Command and Address hold time to CK_t,CK_c referenced to Vih(dc) / Vil(dc) levels	tIH (base)	80	-	73	-	65	-	ps	
Command and Address hold time to CK_t, CK_c referenced to Vref levels	tIH (Vref)	145	-	138	-	130	-	ps	
Control and Address Input pulse width for each	tIPW	385	-	365	-	340	-	ps	
Command and Address Timing									
CAS_n to CAS_n command delay for same bank	tCCD_L	Min = Max (5nCK, 5 ns)						nCK	34
CAS_n to CAS_n command delay for different bank group	tCCD_S	4	-	4	-	4	-	nCK	34
ACTIVATE to ACTIVATE Command delay to different	tRRD_S (2K)	Min = Max (4nCK, 5.3 ns)						nCK	34

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
bank group for 2KB page size									
ACTIVATE to ACTIVATE Command delay to different bank group for 1KB page size	tRRD_S (1K)	Min = Max (4nCK, 3.0 ns)		Min = Max (4nCK, 2.7 ns)		Min = Max (4nCK, 2.5 ns)		nCK	34
ACTIVATE to ACTIVATE Command delay to different bank group for 1/ 2KB page size	tRRD_S (1/2K)	Min = Max (4nCK, 3.0 ns)		Min = Max (4nCK, 2.7 ns)		Min = Max (4nCK, 2.5 ns)		nCK	34
ACTIVATE to ACTIVATE Command delay to same bank group for 2KB page size	tRRD_L (2K)	Min = Max (4nCK, 6.4 ns)		Min = Max (4nCK, 6.4 ns)		Min = Max (4nCK, 6.4 ns)		nCK	34
ACTIVATE to ACTIVATE Command delay to same bank group for 1KB page size	tRRD_L (1K)	Min = Max (4nCK, 4.9 ns)		Min = Max (4nCK, 4.9 ns)		Min = Max (4nCK, 4.9 ns)		nCK	34
ACTIVATE to ACTIVATE Command delay to same bank group for 1/ 2KB page size	tRRD_L (1/2K)	Min = Max (4nCK, 3.0 ns)		Min = Max (4nCK, 2.7 ns)		Min = Max (4nCK, 2.5 ns)		nCK	34
Four activate window for 2KB page size	tFAW_2K	Min = Max (28nCK, 30 ns)		Min = Max (28nCK, 30 ns)		Min = Max (28nCK, 30 ns)		ns	34
Four activate window for 1KB page size	tFAW_1K	Min = Max (20nCK, 21 ns)		Min = Max (20nCK, 21 ns)		Min = Max (20nCK, 21 ns)		ns	34
Four activate window for 1/2KB page size	tFAW_1/2K	Min = Max (16nCK, 12 ns)		Min = Max (16nCK, 10.875 ns)		Min = Max (16nCK, 10 ns)		ns	34
Delay from start of internal Write transaction to internal Read command for different bank group	tWTR_S	Min = Max (2nCK, 2.5 ns)		Min = Max (2nCK, 2.5 ns)		Min = Max (2nCK, 2.5 ns)		ns	1, 2, 34
Delay from start of internal write transaction to internal read command for same bank group	tWTR_L	Min = Max (4nCK, 7.5 ns)		Min = Max (4nCK, 7.5 ns)		Min = Max (4nCK, 7.5 ns)		ns	1, 34
Internal READ Command to PRE - CHARGE Command delay	tRTP	Min = Max (4nCK, 7.5 ns)		Min = Max (4nCK, 7.5 ns)		Min = Max (4nCK, 7.5 ns)		ns	
WRITE recovery time	tWR	15	-	15	-	15	-	ns	1
Write recovery time when CRC and DM are enabled	tWR_CRC_DM	Min = tWR + max(5nCK, 3.75 ns)						ns	1, 28
Delay from start of internal Write transaction to internal	tWTR_S_CRC_DM	Min = tWTR_S + max(5nCK, 3.75 ns)						ns	2,,29, 34

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
read command for different bank group with both CRC and DM enabled									
Delay from start of internal Write transaction to internal read command for same bank group with both CRC and DM enabled	tWTR_L_CRC_DM	Min = tWTR_L + max(5nCK, 3.75 ns)						ns	3, 30, 34
DDL locking time	tDLLK	1024	-	1024	-	1024	-	nCK	
Mode Register Set command cycle time	tMRD	8	-	8	-	8	-	nCK	
Mode Register Set command update delay	tMOD	Min = Max (24nCK, 15 ns)						nCK	
Multi-Purpose Register Recovery Time	tMPRR	1	-	1	-	1	-	nCK	33
Multi-Purpose Register Write Recovery Time	tWR_MPR	Min = tMOD (min) + AL + PL						nCK	
Auto precharge write recovery + precharge time	tDAL (min)	Programmed WR +roundup (tRP / tCK(avg))						nCK	
DQ0 or DQL0 driven to set-up time to first DQS rising edge	tPDA_S	0.5	-	0.5	-	0.5	-	UI	45, 47
DQ0 or DQL0 driven to hold time from last DQS falling edge	tPDA_H	0.5	-	0.5	-	0.5	-	UI	46, 47
CS_n to Command Address Latency									
CS_n to Command Address Latency	tCAL	Max (3nCK, 3.748 ns)	-	Max (3nCK, 3.748 ns)	-	Max (3nCK, 3.748 ns)	-	nCK	
Mode Register Set command cycle time in CAL mode	tMRD_tCAL	tMOD + tCAL	-	tMOD + tCAL	-	tMOD + tCAL	-	nCK	
Mode Register Set update delay in CAL mode	tMOD_tCAL	tMOD + tCAL	-	tMOD + tCAL	-	tMOD + tCAL	-	nCK	
DRAM Data Timing									
DQS_t,DQS_c to DQ skew, per group, per acces	tDQSQ	-	0.18	-	0.19	-	TBD	tCK	13,18,39,49

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
								(avg)/2	
DQ output hold time from DQS_t, DQS_c	tQH	0.74	-	0.72	-	TBD	-	tCK (avg)/2	17,18,39,49
Data Valid Window per device per UI: (tQH - tDQSQ) of each UI on a given DRAM	tDVWd	0.64	-	TBD	-	TBD	-		
Data Valid Window per pin per UI: (tQH - tDQSQ) each UI on a pin of a given DRAM	tDVWp	0.72	-	TBD	-	0.72	-		
DQ low impedance time from CK_t, CK_c	tLZ (DQ)	-310	170	-280	165	-250	160		
DQ high impedance time from CK_t, CK_c	tHZ (DQ)	-	170	-	165	-	160		
Data Strobe Timing									
DQS_t, DQS_c differential READ Preamble (1 clock preamble)	tRPRE	0.9	Note 44	0.9	Note 44	0.9	Note 44	tCK	
DQS_t, DQS_c differential READ Preamble (2 clock preamble)	tRPRE2	1.8	Note 44	1.8	Note 44	1.8	Note 44	tCK	
DQS_t, DQS_c differential READ Postamble	tRPST	0.33	Note 45	0.33	Note 45	0.33	Note 45	tCK	
DQS_t, DQS_c differential output high time	tQSH	0.4	-	0.4	-	0.4	-	tCK	21,39
DQS_t, DQS_c differential output low time	tQSL	0.4	-	0.4	-	0.4	-	tCK	20,39
DQS_t, DQS_c differential WRITE Preamble (1 clock preamble)	tWPRE	0.9	-	0.9	-	0.9	-	tCK	42
DQS_t, DQS_c differential WRITE Preamble (2 clock preamble)	tWPRE2	1.8	-	1.8	-	1.8	-	tCK	43
DQS_t, DQS_c differential WRITE Postamble	tWPST	0.33	-	0.33	-	0.33	-	tCK	
DQS_t and DQS_c low-impedance time (Referenced from RL-1)	tLZ (DQS)	-310	170	-280	165	-250	160	ps	39
DQS_t and DQS_c high-impedance time (Referenced	tHZ (DQS)	-	170	-	165	-	160	ps	39

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
from RL+BL/2)									
DQS_t, DQS_c differential input low pulse width	tDQSL	0.46	0.54	0.46	0.54	0.46	0.54	tCK	
DQS_t, DQS_c differential input high pulse width	tDQSH	0.46	0.54	0.46	0.54	0.46	0.54	tC	
DQS_t, DQS_c rising edge to CK_t, CK_c rising edge (1 clock preamble)	tDQSS	-0.27	0.27	-0.27	0.27	-0.27	0.27	tCK	42
DQS_t, DQS_c rising edge to CK_t, CK_c rising edge (2 clock preamble)	tDQSS2	-0.5	0.5	TBD	TBD	TBD	TBD		43
DQS_t, DQS_c falling edge setup time to CK_t, CK_c rising edge	tDSS	0.18	-	0.18	-	0.18	-		
DQS_t, DQS_c falling edge hold time from CK_t, CK_c rising edge	tDSH	0.18	-	0.18	-	0.18	-	tCK	
DQS_t, DQS_c rising edge output variance window per DRAM	tDQSCKI (DLL On)	-	270	-	265	-	260	ps	37,38,39
DQS_t, DQS_c rising edge output timing location from rising CK_t, CK_c with DDL on mode	tDQSCK	-170	170	-165	165	-160	160	ps	37,38,39
MPSM Timing									
Command path disable delay upon MPSM entry	tMPED	tMOD(min) + tCPDED(min)	-	tMOD(min) + tCPDED(min)	-	tMOD(min) + tCPDED(min)	-	tCK	
Valid clock requirement after MPSM entry	tCKMPE	tMOD(min) + tCPDED(min)	-	tMOD(min) + tCPDED(min)	-	tMOD(min) + tCPDED(min)	-	tCK	
Valid clock requirement before MPSM exit	tCKMPX	tCKSRX(min)	-	tCKSRX(min)	-	tCKSRX(min)	-	tCK	
Exit MPSM to commands not requiring a locked DLL	tXMP	tXS(min)	-	tXS(min)	-	tXS(min)	-	tCK	
Exit MPSM to commands requiring a locked DLL	tXMPDLL	tXMP(min) + tXSDLL(min)	-	tXMP(min) + tXSDLL(min)	-	tXMP(min) + tXSDLL(min)	-	tCK	
CS setup time to CKE	tMPX_S	tISmin + tIHmin	-	tISmin + tIHmin	-	tISmin + tIHmin	-	ns	

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
Calibration Timing									
Power-up and RESET calibration time	tZQinit	1024	-	1024	-	1024	-	nCK	
Normal operation Full calibration time	tZQoper	512	-	512	-	512	-	nCK	
Normal operation Short calibration time	tZQCS	128	-	128	-	128	-	nCK	
Reset/Self Refresh Timing									
Exit Reset from CKE HIGH to a valid command	tXPR	Min = Max(5nCK, tRFC(min) + 10 ns)							
Exit Self Refresh to commands not requiring a locked DLL	tXS	Min = tRFC (min) + 10ns							
SRX to Commands not requiring a locked DLL in Self Refresh ABORT	tXS_ABORT(min)	Min = tRFC4 (min) + 10ns							
Exit Self Refresh to ZQCL,ZQCS and MRS (CL,CWL,WR,RTP and Gear Down)	tXS_FAST(min)	Min = tRFC4 (min) + 10ns							
Exit Self Refresh to commands requiring a locked DLL	tXSDLL	tDLLK (min)	-	tDLLK (min)	-	tDLLK (min)	-		
Minimum CKE low width for Self refresh entry to exit timing	tCKESR	tCKE (min) + 1nCK	-	tCKE (min) + 1nCK	-	tCKE (min) + 1nCK	-		
Minimum CKE low width for Self refresh entry to exit timing with CA Parity enabled	tCKESR_PAR	tCKE (min) + 1nCK + PL	-	tCKE (min) + 1nCK + PL	-	tCKE (min) + 1nCK + PL	-		
Valid Clock Requirement after Self Refresh Entry (SRE) or Power-Down Entry (PDE)	tCKSRE	Max (5nCK, 10 ns)	-	Max (5nCK, 10 ns)	-	Max (5nCK, 10 ns)	-		
Valid Clock Requirement after Self Refresh Entry (SRE) or Power-Down when CA Parity is enabled	Tcksre_PAR	Max (5nCK, 10 ns) + PL	-	Max (5nCK, 10 ns) + PL	-	Max (5nCK, 10 ns) + PL	-		
Valid Clock Requirement before Self Refresh Exit (SRX) or Power Down Exit (PDX) or Reset Exit	tCKSRX	Max (5nCK, 10 ns)	-	Max (5nCK, 10 ns)	-	Max (5nCK, 10 ns)	-		

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note	
Parameter	Symbol	Min	Max	Min	Max	Min	Max			
Power Down Timing										
Exit Power Down with DLL on to any valid command, Exit Precharge Power Down with DLL frozen to	tXP	Max (4nCK, 6ns)	-	Max (4nCK, 6ns)	-	Max (4nCK, 6ns)	-			
CKE minimum pulse width	tCKE	Max (3nCK, 5ns)	-	Max (3nCK, 5ns)	-	Max (3nCK, 5ns)	-		31,32	
Command pass disable delay	tCPDED	4	-	4	-	4	-	nCK		
Power Down Entry to Exit Timing	tPD	tCKE (min)	9*tREFI	tCKE (min)	9*tREFI	tCKE (min)	9*tREFI		6	
Timing of ACT command to Power Down entry	tACTPDEN	2	-	2	-	2	-	nCK	7	
Timing of PRE or PREA command to Power Down	tPRPDEN	2	-	2	-	2	-	nCK	7	
Timing of RD/RDA command to Power Down entry	tRDPDEN	Min = RL + 4 + 1							nCK	
Timing of WR command to Power Down entry (BL8OTF, BL4MRS, BC4OTF)	tWRPDEN	Min = WL + 4 + (tWR/tCK(avg))							nCK	4
Timing of WRA command to Power Down entry (BL8OTF, BL8MRS, BC4OTF)	tWRAPDEN	Min = WL + 4 + WR + 1							nCK	5
Timing of WR command to Power Down entry (BC4MRS)	tWRP-BC4DEN	Min = WL + 2 + (tWR/tCK(avg))							nCK	4
Timing of WRA command to Power Down entry (BC4MRS)	tWRAP-BC4DEN	Min = WL + 2 + WR + 1							nCK	5
Timing of REF command to Power Down entry	tREFPDEN	2	-	2	-	2	-	nCK	7	
Timing of MRS command to Power Down entry	tMRSPDEN	tMOD (min)	-	tMOD (min)	-	tMOD (min)	-			
Mode Register Set command cycle time in PDA mode	tMRD_PDA	Max (16nCK, 10ns)	-	Max (16nCK, 10ns)	-	Max (16nCK, 10ns)	-			
Mode Register Set command update delay in PDA mode	tMOD_PDA	tMOD		tMOD		tMOD				
Asynchronous RTT turn-on delay (Power-Down with DLL)	tAONAS	1	9	1	9	1	9	ns		

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
frozen)									
Asynchronous RTT turn-off delay (Power-Down with DLL frozen)	tAOFAS	1	9	1	9	1	9	ns	
RTT dynamic change skew	tADC	0.28	0.72	0.26	0.74	0.26	0.74	tCK(avg)	
First DQS_t/DQS_n rising edge after write leveling mode is programmed	tWLMRD	40	-	40	-	40	-	nCK	12
DQS_t/DQS_n delay after write leveling mode is programmed	tWLDQSEN	25	-	25	-	25	-	nCK	12
Write leveling setup time from rising CK_t, CK_c crossing to rising DQS_t/DQS_n crossing	tWLS	0.13	-	0.13	-	0.13	-	tCK(avg)	
Write leveling hold time from rising DQS_t/DQS_n crossing to rising CK_t, CK_crossing	tWLH	0.13	-	0.13	-	0.13	-	tCK(avg)	
Write leveling output delay	tWLO	0	0.95	0	0.95	0	9.5	ns	
Write leveling output error	tWLOE	-	2	-	2	-	2	ns	
CA Parity Timing									
Commands not guaranteed to be executed during this time	tPAR_UNKNOWN	-	PL	-	PL	-	PL	nCK	
Delay from errant command to ALERT_n assertion	tPAR_ALERT_ON	-	PL + 6ns	-	PL + 6ns	-	PL + 6ns	nCK	
Pulse width of ALERT_n signal when asserted	tPAR_ALERT_PW	8	160	8	176	9	192	nCK	
Time from when Alert is asserted till controller must start providing DES commands in Persistent CA parity mode	tPAR_ALERT_RSP	-	71	-	78	-	85	nCK	
Parity Latency	PL	5		6		6		nCK	
CRC error to ALERT_n latency	tCRC_ALERT	3	13	3	13	3	13	ns	
CRC ALERT_n pulse width	CRC_ALERT_PW	6	10	6	10	6	10	nCK	
Gear Down Timing									

Speed		DDR4-2666		DDR4-2933		DDR4-3200		Unit	Note
Parameter	Symbol	Min	Max	Min	Max	Min	Max		
Exit RESET from CKE HIGH to a valid MRS geardown (T2/Reset)	tXPR_GEAR	tXPR	-	tXPR	-	tXPR	-		
CKE High Assert to Gear Down Enable time (T2/CKE)	tXS_GEAR	tXS	-	tXS	-	tXS	-		
MRS command to Sync pulse time(T3)	tSYNC_GEAR	tMOD +4nCK	-	tMOD +4nCK	-	tMOD +4nCK	-		27
Sync pulse to First valid command(T4)	tCMD_GEAR	tMOD	-	tMOD	-	tMOD	-		27
Geardown setup time	tGEAR setup	2	-		-	2	-	nCK	
Geardown hold time	tGEAR hold	2	-		-	2	-	nCK	
tREFI									
tRFC1 (min)	4Gb	260	-	260	-	260	-	ns	34
tRFC2 (min)	4Gb	160	-	160	-	160	-	ns	34
tRFC4 (min)	4Gb	110	-	110	-	110	-	ns	34

Note:

1. Start of internal Write transaction is defined as follows:
 - For BL8 (Fixed by MRS and on-the-fly): Rising clock edge 4 clock cycles after WL.
 - For BC4 (on-the-fly): Rising clock edge 4 clock cycles after WL.
 - For BC4 (Fixed by MRS): Rising clock edge 2 clock cycles after WL.
2. A separate timing parameter will cover the delay from Write to Read when CRC and DM are simultaneously enabled.
3. Commands requiring a locked DLL are Read (and Read Auto Precharge) and synchronous ODT commands.
4. tWR is defined in ns, for calculation of tWRPDEN it is necessary to round up tWR/tCK following rounding algorithm defined in [Section 11.5](#).
5. WR in clock cycles as programmed in MR0.
6. tREFI depends on T_{OPER}.
7. CKE is allowed to be registered low while operations such as row activation, precharge, autoprecharge or refresh are in progress, but power-down IDD spec will not be applied until finishing those operations.
8. For these parameters, the DDR4 SDRAM device supports $t_{nPARAM}[nCK]=RU\{t_{PARAM}[ns]/t_{CK}(avg)[ns]\}$, which is in clock cycles assuming all input clock jitter specifications are satisfied.
9. When CRC and DM are both enabled, tWR_CRC_DM is used in place of tWR.

10. When CRC and DM are both enabled, tWTR_S_CRC_DM is used in place of tWTR_S.
11. When CRC and DM are both enabled, tWTR_L_CRC_DM is used in place of tWTR_L.
12. The max value are system dependent.
13. DQ to DQS total timing per group where the total includes the sum of deterministic and random timing terms for a specified BER. BER spec and measurement.
14. The deterministic component of the total timing. Measurement method TBD.
15. DQ to DQ static offset relative to strobe per group. Measurement method TBD.
16. This parameter will be characterized and guaranteed by design.
17. When the device is operated with the input clock jitter, this parameter needs to be derated by the actual tjit(per)_total of the input clock. (output deratings are relative to the SDRAM input clock). Example TBD.
18. DRAM DBI mode is off.
19. DRAM DBI mode is enabled. Applicable to x8 and x16 DRAM only.
20. tQSL describes the instantaneous differential output low pulse width on DQS_t - DQS_c, as measured from on falling edge to the next consecutive rising edge.
21. tQSL describes the instantaneous differential output high pulse width on DQS_t - DQS_c, as measured from on falling edge to the next consecutive rising edge.
22. There is no maximum cycle time limit besides the need to satisfy the refresh interval tREFI.
23. tCH(abs) is the absolute instantaneous clock high pulse width, as measured from one rising edge to the following falling edge.
24. tCL(abs) is the absolute instantaneous clock low pulse width, as measured from one falling edge to the following rising edge.
25. Total jitter includes the sum of deterministic and random jitter terms for a specified BER. BER target and measurement method are TBD.
26. The deterministic jitter component out of the total jitter. This parameter is characterized and guaranteed by design.
27. This parameter has to be even number of clocks.
28. When CRC and DM are both enabled, tWR_CRC_DM is used in place of tWR.
29. When CRC and DM are both enabled tWTR_S_CRC_DM is used in place of tWTR_S.
30. When CRC and DM are both enabled tWTR_L_CRC_DM is used in place of tWTR_L.
31. After CKE is registered Low, CKE signal level shall be maintained below VILDC for tCKE specification (Low pulse width).
32. After CKE is registered High, CKE signal level shall be maintained above VIHDC for tCKE specification (High pulse width).
33. Defined between end of MPR Read burst and MRS which reloads MPR or disables MPR function.
34. Parameters apply from tCK(avg)min to tCK(avg)max at all standard JEDEC clock period values as stated in the "Speed Bin".
35. This parameter must keep consistency with "Speed Bin".

36. DDR4-1600 AC timing apply if DRAM operates at lower than 1600 MT/s data rate. $UI=tCK(avg)min/2$.
37. Applied when DRAM is in DLL ON mode.
38. Assume no jitter on input clock signals to the DRAM.
39. Value is only valid for $RONNOM = 34\Omega$.
40. 1tCK toggle mode with setting MR4:A11 to 0.
41. 2tCK toggle mode with setting MR4:A11 to 1, which is valid for DDR4-2400/2666/3200 speed grade.
42. 1tCK mode with setting MR4:A12 to 0.
43. 2tCK mode with setting MR4:A12 to 1, which is valid for DDR4-2400/2666/3200 speed grade.
44. The maximum read preamble is bounded by $tLZ(DQS)min$ on the left side and $tDQSCK(max)$ on the right side. See "Clock to Data Strobe Relationship" of DDR4 Operation Guide Boundary of DQS Low-Z occur one cycle earlier in 2tCK toggle mode which is illustrated in "Read Preamble".
45. DQ falling signal middle-point of transferring from High to Low to first rising edge of DQS diff-signal cross-point.
46. Last falling edge of DQS diff-signal cross-point to DQ rising signal middle-point of transferring from Low to High
47. V_{refDQ} value must be set to either its midpoint or $V_{cent_DQ}(midpoint)$ in order to capture DQ0 or DQL0 low level for entering PDA mode.
48. The maximum read postamble is bound by $tDQSCK(min)$ plus $tQSH(min)$ on the left side and $tHZ(DQS)max$ on the right side. See "Clock to Data Strobe Relationship" of DDR4 Operation Guide.
49. Reference level of DQ output signal is specified with a midpoint as a widest part of Output signal eye which should be approximately $0.7 * VDDQ$ as a center level of the static single-ended output peak-to-peak swing with a driver impedance of 34 ohms and an effective test load of 50 ohms to $V_{TT} = VDDQ$.
50. For MR7 commands, the minimum delay to a subsequent non-MRS command is 5nCK.

9.5 Rounding Algorithms

Software algorithms for calculation of timing parameters are subject to rounding errors from many sources. For example, a system may use a memory clock with a nominal frequency of 933.33 MHz, or a clock period of 1.0714 ns. Similarly, a system with a memory clock frequency of 1066.66 MHz yields mathematically a clock period of 0.9375 ns. In most cases, it is impossible to express all digits after the decimal point exactly, and rounding must be done because the DDR4 SDRAM specification establishes a minimum granularity for timing parameters of 1 ps.

Rules for rounding must be defined to allow optimization of device performance without violating device parameters. These algorithms rely on results that are within correction factors on device testing and specification to avoid losing performance due to rounding errors. These rules are:

- Clock periods such as tCKAVGmin are defined to 1 ps of accuracy; for example, 0.9375... ns is defined as 937 ps and 1.0714 ns is defined as 1071 ps.
- Using real math, parameters like tAmin, tRCDmin, etc. which are programmed in systems in numbers of clocks (nCK) but expressed in units of time (in ns) are divided by the clock period (in ns) yielding a unitless ratio, a correction factor of 2.5% is subtracted, then the result is set to the next higher integer number of clocks:

$$nCK = \text{ceiling} [(\text{parameter_in_ns} / \text{application_tCK_in_ns}) - 0.025]$$
- Alternatively, programmers may prefer to use integer math instead of real math by expressing timing in ps, scaling the desired parameter value by 1000, dividing by the application clock period, adding an inverse correction factor of 97.4%, dividing the result by 1000, then truncating down to the next lower integer value:

$$nCK = \text{truncate} [\{ (\text{parameter_in_ps} \times 1000) / (\text{application_tCK_in_ps}) + 974 \} / 1000]$$
- Either algorithm yields identical results. In case of conflict between results, the preferred algorithm is the integer math algorithm.
- This algorithm applies to all timing parameters documented in a Serial Presence Detect (SPD) when converting from ns to nCK. Other timing parameters may use a simpler algorithm:

$$nCK = \text{ceiling} (\text{parameter_in_ns} \div \text{application_tCK_in_ns}).$$

9.6 The DQ Input Receiver Compliance Mask for Voltage and Timing

The DQ input receiver compliance mask for voltage and timing is shown in Figure 11-2 below. The receiver mask (Rx Mask) defines area the input signal must not encroach in order for the DRAM input receiver to be able to successfully capture a valid input signal with BER of 1e-16; Any input signal encroaching within the Rx Mask is subject to being invalid data. The Rx Mask is the receiver property for each DQ input pin and it is not the valid data-eye.

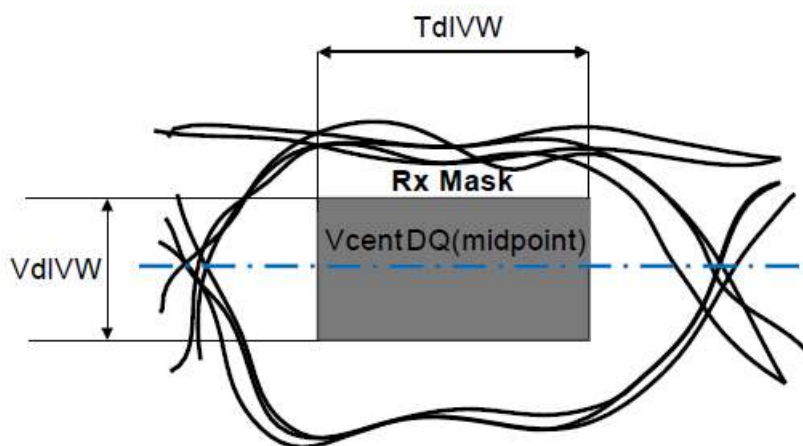


Figure 9-2. DQ Receiver(Rx) Compliance mask

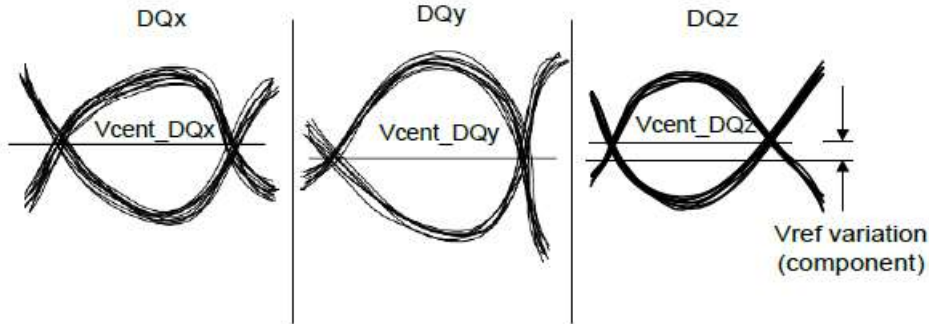


Figure 9-3. Across Pin Vref DQ Voltage Variation

The Vref_DQ voltage is an internal reference voltage level that shall be set to the properly trained setting, which is generally Vcent_DQ(midpoint), in order to have valid Rx Mask values.

Vcent_DQ(pin avg) is defined as the midpoint between the largest Vref_DQ voltage level and the smallest Vref_DQ voltage level across all DQ pins for a given DRAM component. Each DQ pin Vref level is defined by the center, i.e. widest opening, of the cumulative data input eye as depicted in Figure 11-3. This clarifies that any DRAM component level variation must be accounted for within the DRAM Rx mask. The component level Vref will be set by the system to account for RON and ODT settings.

Figure 9-4. DQS to DQ and DQ to DQ Timings at DRAM Balls

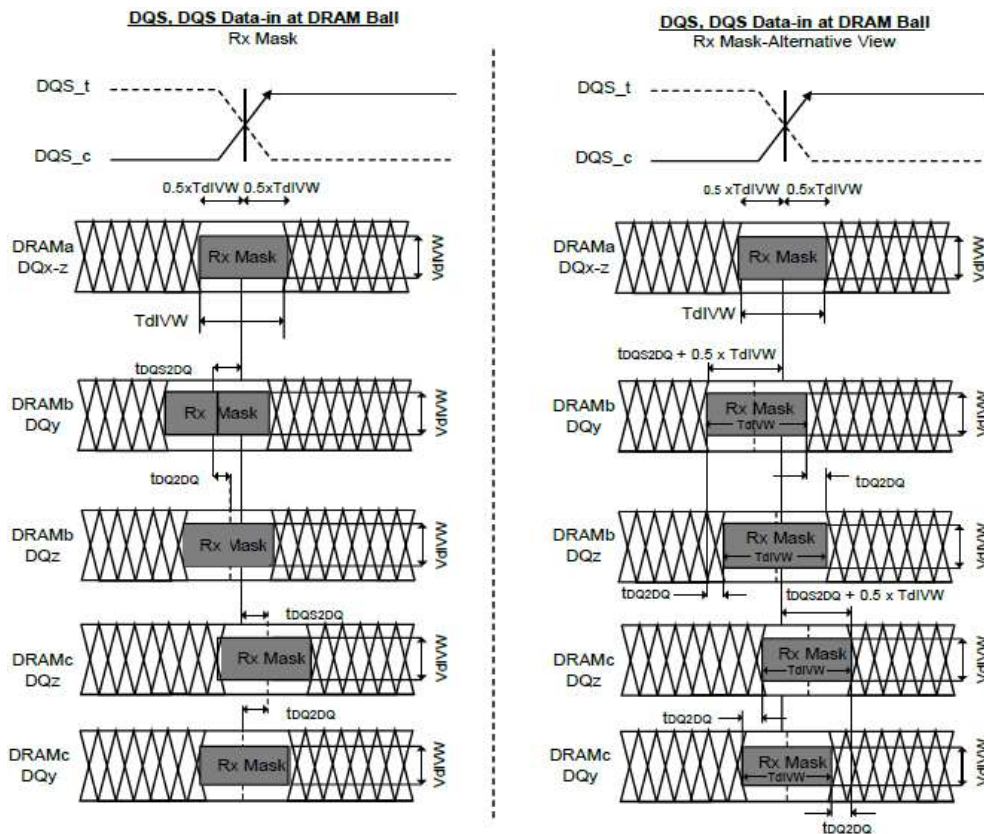


Figure 11-4. DQS to DQ and DQ to DQ Timings at DRAM Balls

Note:

- DQx represents an optimally centered mask.
- DQy represents earliest valid mask.
- DQz represents lasted valid mask.

Note:

- DRAMa represents a DRAM without any DQS/DQ skews.
- DRAMB represents a DRAM with early skews (negative tba2DQ).
- DRAMc represents a DRAM with delayed skews (positive tba2DQ).

Note:

1. Figures show skew allowed between DRAM to DRAM and DQ to DQ for a DRAM. Signals assume data centered aligned at DRAM Latch.
2. TdiPW is not shown; composite data-eyes shown would violate TdiPW.
3. VCENT DQ(midpoint) is not shown but is assumed to be midpoint of VdiVW.

All of the timing term in are measured at the VdiVW voltage levels centered around

Vcent_DQ(midpoint) and are referenced to the DQS_t/DQS_c center aligned to the DQ per pin.

The rising edge slew rates are defined by srr1 and srr2. The slew rate measurement points for a rising edge are shown in Figure 11-4 below: A low to high transition tr1 is measured from 0.5*VdiVW(max) below Vcent_DQ(midpoint) to the last transition through 0.5*VdiVW(max) above Vcent_DQ(midpoint) to the first transition through the 0.5*VIHL_AC(min) above Vcent_DQ(midpoint).

Rising edge slew rate equations:

- $srr1 = VdiVW(max) / tr1$
- $srr2 = (VIHL_AC(min) - VdiVW(max)) / (2*tr2)$

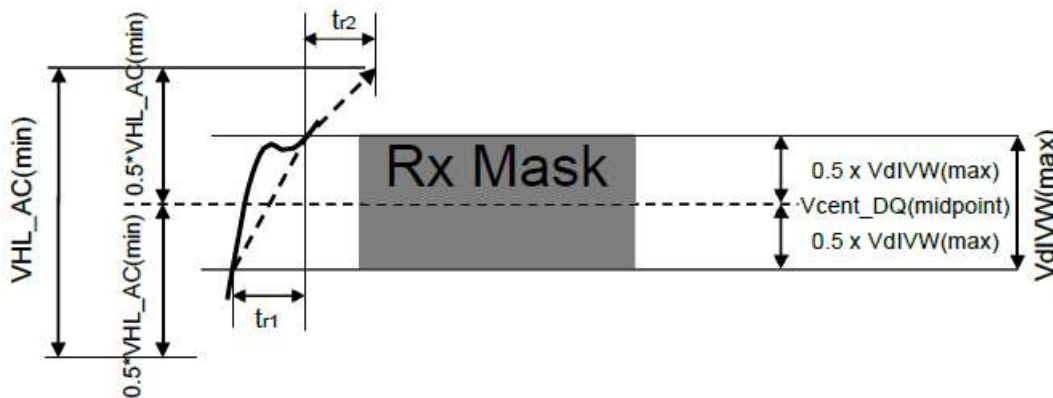


Figure 9-4. Slew Rate Conditions for Rising Transition

The falling edge slew rates are defined by srf1 and srf2. The slew rate measurement points for a falling edge are shown in Figure 11-5 below: A high to low transition tf1 is measured from 0.5*VdiVW(max) above Vcent_DQ(midpoint) to the last transition through 0.5*VdiVW(max) below Vcent_DQ(midpoint) while tf2 is measured from the last transition through 0.5*VdiVW(max) below Vcent_DQ(midpoint) to the first transition through the 0.5*VIHL_AC(min) below Vcent_DQ(pin mid).

Falling edge slew rate equations:

- $srf1 = VdiVW(max) / tf1$
- $srf2 = (VIHL_AC(min) - VdiVW(max)) / (2*tf2)$

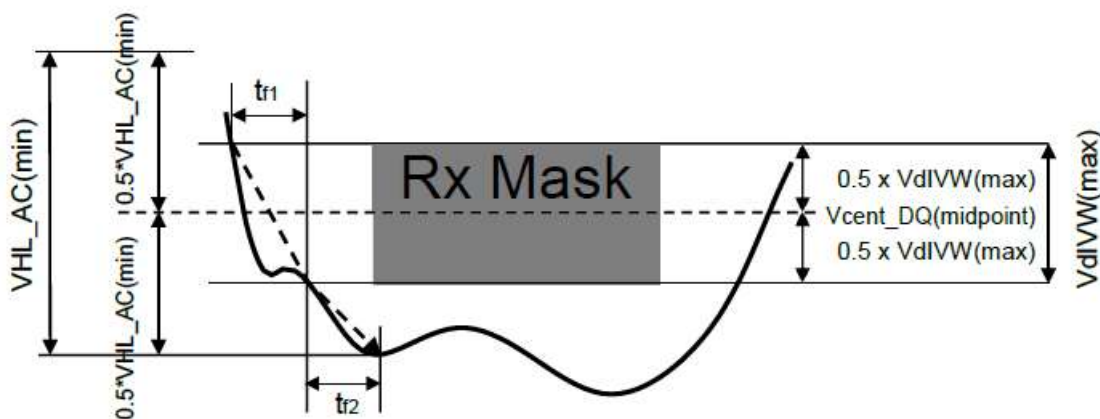


Figure 9-5. Slew Rate Conditions for Falling Transition

Table 9-4. DRAM DQS in Receive Mode; * UI=tck(avg)min/2

Symbol	Parameter	DDR4-3200		Unit	Note
		Min	Max		
VdIVW	Rx Mask voltage – pK-pK	-	112	mV	1,2,10
TdIVW	Rx timing window	-	0.22	UI*	1,2,10
VIHL_AC	DQ AC input swing pk-pk	150	-	mV	3,4,10
TdIPW	DQ input pulse width	0.58	-	UI*	5,10
tDQS2DQ	Rx Mask DQS to DQ offset	-0.19	0.19	UI*	6,10
tDQ2DQ	Rx Mask DQ to DQ offset	-	0.105	UI*	7
srr1	Input Slew Rate over VdIVW if tCK >=0.937ns	1	9	V/ns	8,10,
srf1	Input Slew Rate over VdIVW if 0.937ns > tCK >= 0.625ns	1.25	9	V/ns	8,10,
srr2	Rising Input Slew Rate over 1/2 VIHL_AC	0.2*srr1	9	V/ns	9,10,
srf2	Falling Input Slew Rate over 1/2 VIHL_AC	0.2*srr1	9	V/ns	9,10,

Note:

1. Data Rx mask voltage and timing total input valid window where VdIVW is centered around Vcent_DQ (midpoint) after VrefDQ training is completed. The data Rx mask is applied per bit and should include voltage and temperature drift terms. The input buffer design specification is to achieve at least a BER = e-16 when the Rx mask is not violated. The BER will be characterized and extrapolated if necessary using a dual dirac method from a higher BER(tbd)
2. Defined over the DQ internal Vref range 1.
3. Overshoot and Undershoot Specifications see “AC Overshoot Specification for Data, Strobe and Mask”
4. DQ input pulse signal swing into the receiver must meet or exceed VIHL AC(min). VIHL_AC(min) is to be achieved on an UI basis when a rising and falling edge occur in the same UI, i.e. a valid TdiPW.

5. DQ minimum input pulse width defined at the V_{cent_DQ} (midpoint).
6. DQS to DQ offset is skew between DQS and DQs within a nibble (x4) or word (x8, x16) at the DDR4 SDRAM balls over process, voltage, and temperature.
7. DQ to DQ offset is skew between DQs within a nibble (x4) or word (x8, x16) at the DDR4 SDRAM balls for a given component over process, voltage, and temperature.
8. Input slew rate over V_{dIVW} Mask centered at V_{cent_DQ} (midpoint). Slowest DQ slew rate to fastest DQ slew rate per transition edge must be within 1.7 V/ns of each other.
9. Input slew rate between V_{dIVW} Mask edge and V_{IHL_AC} (min) points.
10. All Rx Mask specifications must be satisfied for each UI. For example, if the minimum input pulse width is violated when satisfying $T_{dIVW}(\min)$, $V_{dIVW}(\max)$, and minimum slew rate limits, then either $T_{dIVW}(\min)$ or minimum slew rates would have to be increased to the point where the minimum input pulse width would no longer be violated.

9.7 Command, Control, and Address Setup, Hold, and Derating

The total tIS (setup time) and tIH (hold time) required is calculated to account for slew rate variation by adding the data sheet tIS (base) values, the VIL(AC)/VIH(AC) points, and tIH (base) values, the VIL(DC)/VIH(DC) points; to the ΔtIS and ΔtIH derating values, respectively. The base values are derived with single-end signals at 1V/ns and differential clock at 2V/ns. Example: tIS (total setup time) = tIS (base) + ΔtIS. For a valid transition, the input signal has to remain above/ below VIH(AC)/VIL(AC) for the time defined by tVAC.

Although the total setup time for slow slew rates might be negative (for example, a valid input signal will not have reached VIH(AC)/VIL(AC) at the time of the rising clock transition), a valid input signal is still required to complete the transition and to reach VIH(AC)/VIL(AC). For slew rates that fall between the values listed in derating tables, the derating values may be obtained by linear interpolation.

Setup (tIS) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of VIL(DC)max and the first crossing of VIH(AC)min that does not ring back below VIH(DC)min. Setup (tIS) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of VIH(DC)min and the first crossing of VIL(AC)max that does not ring back above VIL(DC)max. Hold (tIH) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of VIL(DC)max and the first crossing of VIH(AC)min that does not ring back below VIH(DC)min.

Hold (tIH) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of VIH(DC)min and the first crossing of VIL(AC)min that does not ring back above VIL(DC)max.

Table 9-5. Command, Address, Control Setup and Hold Values

DDR4	2666	2933	3200	Unit	Reference
tIS(base, AC100)	-	-	-	ps	VIH/L(ac)
tIH(base, DC75)	-	-	-	ps	VIH/L(dc)
tIS(base, AC90)	55	TBD	40	ps	VIH/L(ac)
tIH(base, DC65)	80	TBD	65	ps	VIH/L(dc)
tIS/Tih@ VREF	145	TBD	130	ps	

Note:

1. Base ac/dc referenced for 1 V/ns slew rate and 2 V/ns clock slew rate.
2. Values listed are referenced only; applicable limits are defined elsewhere.

Table 9-6. Command, Address, Control Input Voltage Values

DDR4	2666	2933	3200	Unit	Reference
VIH. CA(AC)min	90	TBD	90	mV	VIH/L(ac)
VIH. CA(DC)min	65	TBD	65	mV	VIH/L(dc)
VIL. CA(DC)max	-65	TBD	-65	mV	VIH/L(ac)
VIL. CA(AC)max	-90	TBD	-90	mV	VIH/L(dc)

Note:

1. Command, Address, Control input levels relative to VREFCA.
2. Values listed are referenced only; applicable limits are defined elsewhere.

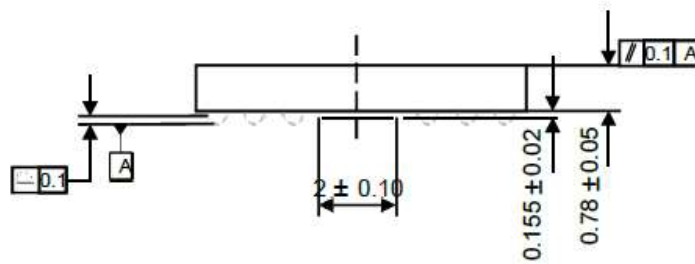
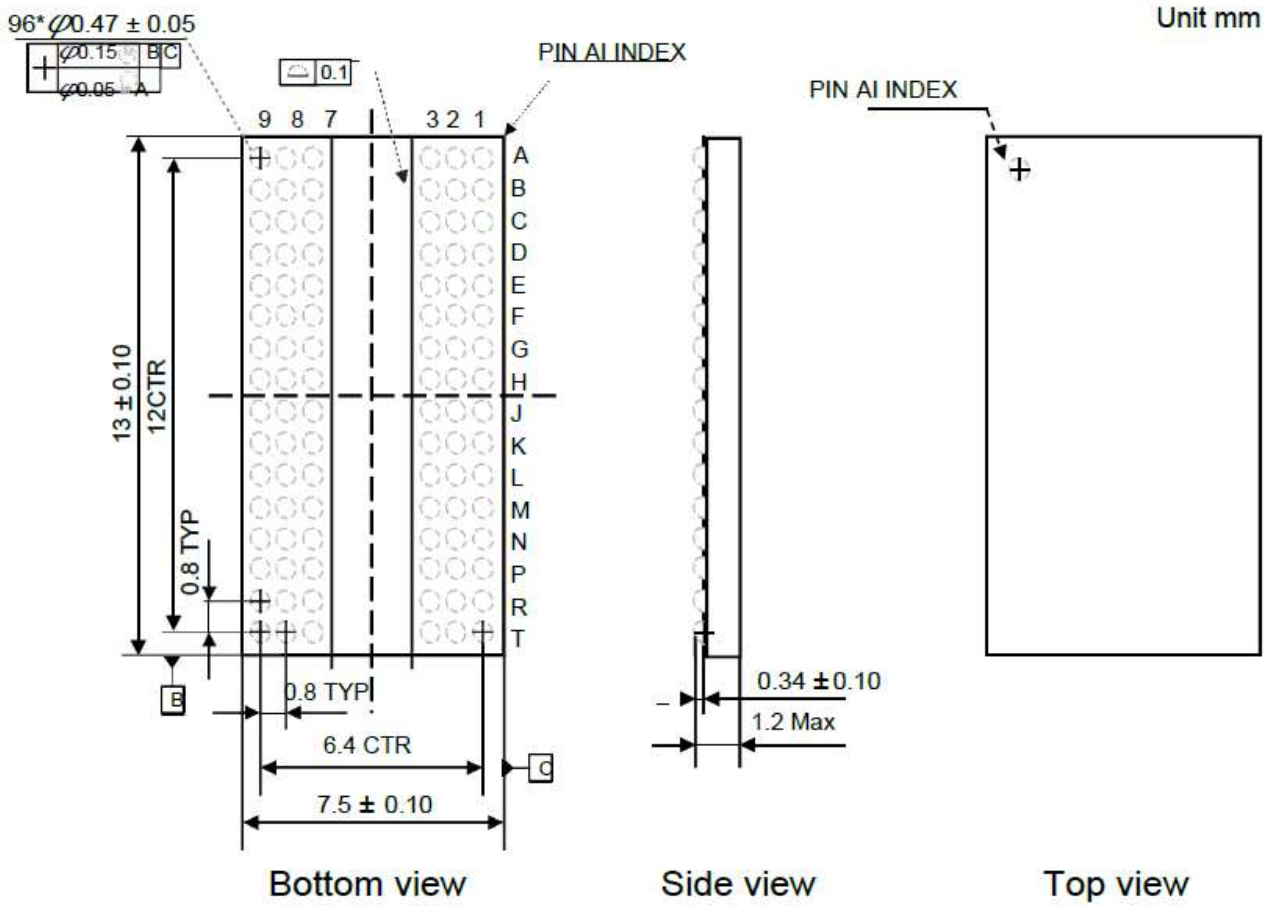
Table 9-7. Derating Values DDR4-3200 tIS/tIH-ac/dc Based

		△ tIS, △ tIH derating in [ps] AC/DC															
		CK_t, CK_c Differential Slew Rate															
		10.0 V/ns		8.0 V/ns		6.0 V/ns		4.0 V/ns		3.0 V/ns		2.0 V/ns		1.5 V/ns		1.0 V/ns	
		tIS	tIH	tIS	tIH	tIS	tIH	tIS	tIH	tIS	tIH	tIS	tIH	tIS	tIH	tIS	tIH
ADDR, CNTL Input Slew Rate V/ns	7	68	47	69	47	70	48	72	50	73	52	77	56	85	63	100	78
	6	66	45	67	46	68	47	69	49	71	50	75	54	83	62	98	77
	5	63	43	64	44	65	45	66	46	68	48	72	52	80	60	95	75
	4	59	40	59	40	60	41	62	43	64	45	68	49	75	56	90	71
	3	51	34	52	35	53	36	54	38	56	40	60	43	68	51	83	66
	2	36	24	37	24	38	25	39	27	41	29	45	33	53	40	68	55
	1.5	21	13	22	13	23	14	24	16	26	18	30	22	38	29	53	44
	1	-9	-9	-8	-8	-8	-8	-6	-6	-4	-4	0	0	8	8	23	23
	0.9	-15	-13	-15	-12	-14	-11	-12	-9	-10	-7	-6	-4	1	4	16	19
	0.8	-23	-17	-23	-17	-22	-16	-20	-14	-18	-12	-14	-8	-7	-1	8	14
	0.7	-34	-23	-33	-22	-32	-21	-30	-20	-28	-18	-25	-14	-17	-6	-2	9
	0.6	-47	-31	-47	-30	-46	-29	-44	-27	-42	-25	-38	-22	-31	-14	-16	1
	0.5	-67	-42	-66	-41	-65	-40	-63	-38	-61	-36	-58	-33	-50	-25	-35	-10
0.4	-95	-58	-95	-57	-94	-56	-92	-54	-90	-53	-86	-49	-79	-41	-64	-26	

Note:

1. VIH/L(ac) = ± tbd mV, VIH/L(dc) = ± tbd mV; relative to VREFCA.

10. PACKAGE DIMENSION (96Ball FBGA, 7.5x13x1.2mm)



11. PART NUMBER LOGIC

